

# OCTAL E1 SHORT HAUL LINE INTERFACE UNIT

IDT82V2058

## **FEATURES**

- Fully integrated octal E1 short haul line interface which supports 120  $\Omega$  E1 twisted pair and 75  $\Omega$  E1 coaxial applications
- Selectable Single Rail mode or Dual Rail mode and AMI or HDB3 encoder/decoder
- Built-in transmit pre-equalization meets G.703
- Selectable transmit/receive jitter attenuator meets ETSI CTR12/ 13, ITU G.736, G.742 and G.823 specifications
- SONET/SDH optimized jitter attenuator meets ITU G.783 mapping jitter specification
- ◆ Digital/Analog LOS detector meets ITU G.775 and ETS 300 233
- ITU G.772 non-intrusive monitoring for in-service testing for any one of channel 1 to channel 7

- ◆ Low impedance transmit drivers with high-Z
- Selectable hardware and parallel/serial host interface
- ◆ Local and Remote Loopback test functions
- Hitless Protection Switching (HPS) for 1 to 1 protection without relays
- JTAG boundary scan for board test
- 3.3 V supply with 5 V tolerant I/O
- Low power consumption
- ◆ Operating temperature range: -40°C to +85°C
- Available in 144-pin Thin Quad Flat Pack (TQFP) and 160-pin Plastic Ball Grid Array (PBGA) packages Green package options available

# **FUNCTIONAL BLOCK DIAGRAM**

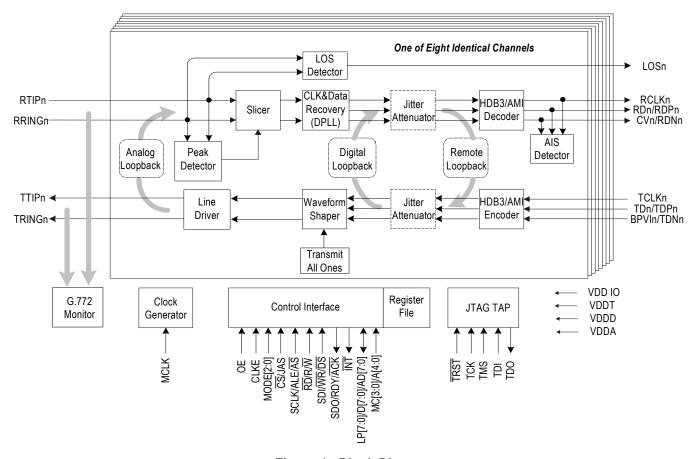


Figure-1 Block Diagram

January 21, 2010

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#### DESCRIPTION

The IDT82V2058 is a single chip, 8-channel E1 short haul PCM transceiver with a reference clock of 2.048 MHz. The IDT82V2058 contains 8 transmitters and 8 receivers.

All the receivers and transmitters can be programmed to work either in Single Rail mode or Dual Rail mode. HDB3 or AMI encoder/decoder is selectable in Single Rail mode. Pre-encoded transmit data in NRZ format can be accepted when the device is configured in Dual Rail mode. The receivers perform clock and data recovery by using integrated digital phase-locked loop. As an option, the raw sliced data (no retiming) can be output on the receive data pins. Transmit equalization is implemented with low-impedance output drivers that provide shaped waveforms to the transformer, guaranteeing template conformance.

A jitter attenuator is integrated in the IDT82V2058 and can be switched into either the transmit path or the receive path for all channels. The jitter attenuation performance meets ETSI CTR12/13, ITU G.736, G.742 and G.823 specifications.

The IDT82V2058 offers hardware control mode and software control mode. Software control mode works with either serial host interface or parallel host interface. The latter works via an Intel/Motorola compatible 8-bit parallel interface for both multiplexed or non-multiplexed applications. Hardware control mode uses multiplexed pins to select different operation modes when the host interface is not available to the device.

The IDT82V2058 also provides loopback and JTAG boundary scan testing functions. Using the integrated monitoring function, the IDT82V2058 can be configured as a 7-channel transceiver with non-intrusive protected monitoring points.

The IDT82V2058 can be used for SDH/SONET multiplexers, central office or PBX, digital access cross connects, digital radio base stations, remote wireless modules and microwave transmission systems.

# PIN CONFIGURATIONS

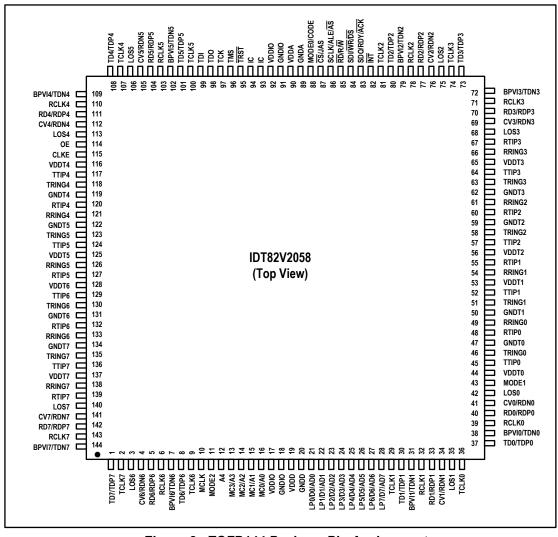


Figure-2 TQFP144 Package Pin Assignment

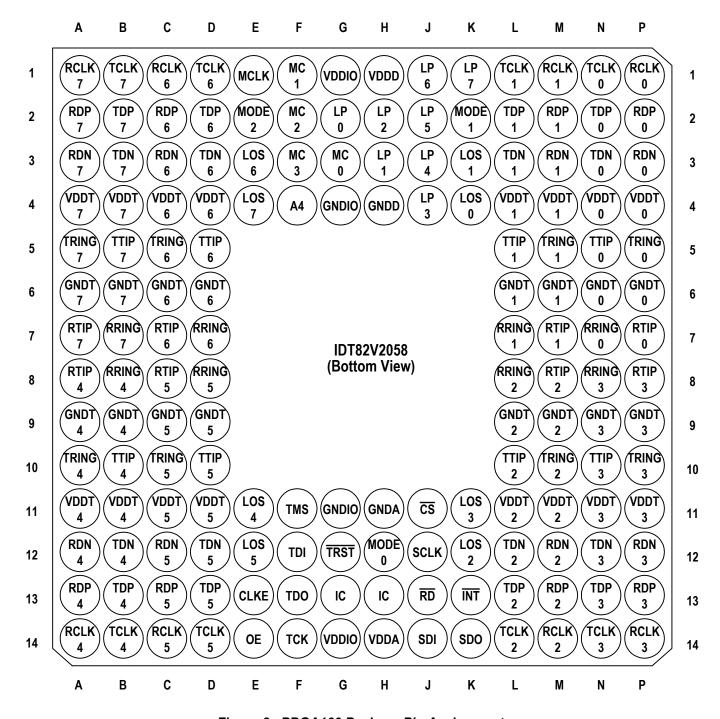


Figure-3 PBGA160 Package Pin Assignment

# 1 PIN DESCRIPTION

Table-1 Pin Description

Name	Туре	Pin	No.	Description			
Name	TQFP144 PBG		PBGA160	Description			
			1	Transmit and Receive Line Interface			
TTIP0 TTIP1 TTIP2		45 52 57	N5 L5 L10				
TTIP3 TTIP4 TTIP5 TTIP6 TTIP7	Analog	64 117 124 129 136	N10 B10 D10 D5 B5	TTIPn/TRINGn: Transmit Bipolar Tip/Ring for Channel 0~7 These pins are the differential line driver outputs. They will be in high-Z if pin OE is low or the correspond-			
TRINGO TRING1 TRING2 TRING3 TRING4 TRING5 TRING6 TRING7	Output	naiog	P5 M5 M10 P10 A10 C10 C5 A5	ing pin TCLKn is low (pin OE is global control, while pin TCLKn is per-channel control). In host mode, each pin can be in high-Z by programming a '1' to the corresponding bit in register <b>OE</b> <sup>(1)</sup> .			
RTIP0 RTIP1 RTIP2 RTIP3 RTIP4 RTIP5 RTIP6 RTIP7  RRING0 RRING1 RRING2 RRING3 RRING4 RRING5 RRING6 RRING5 RRING6 RRING7	Analog Input	48 55 60 67 120 127 132 139 49 54 61 66 121 126 133 138	P7 M7 M8 P8 A8 C8 C7 A7 N7 L7 L8 N8 B8 D8 D7 B7	RTIPn/RRINGn: Receive Bipolar Tip/Ring for Channel 0~7 These pins are the differential line receiver inputs.			

<sup>&</sup>lt;sup>1</sup>. Register name is indicated by bold capital letter. For example, **OE** indicates Output Enable Register.

Table-1 Pin Description (Continued)

Nama	Tuna	Pin	No.				Docorintian	
Name	Type	TQFP144	PBGA160	1			Description	
		•	•	Transmit a	nd Receive Dig	tal Data Inter	face	
TD0/TDP0 TD1/TDP1 TD2/TDP2 TD3/TDP3 TD4/TDP4 TD5/TDP5 TD6/TDP6 TD7/TDP7	ı	37 30 80 73 108 101 8	N2 L2 L13 N13 B13 D13 D2 B2	BPVIn: Bipolar Violation Insertion for Channel 0~7  Bipolar violation insertion is available in Single Rail mode 2 (see Table-2 on page 13 and Table-3 on page 14) with AMI enabled. A low-to-high transition on this pin will make the next logic one to be transmitted on TDn the same polarity as the previous pulse, and violate the AMI rule. This is for testing.  TDPn/TDNn: Positive/Negative Transmit Data for Channel 0~7  When the device is in Dual Rail Mode, the NRZ data to be transmitted for positive/negative pulse is input on this pin. Data on TDPn/TDNn are campled on the folling edges of TCLKe. The line code is dual rail.				
BPVI0/TDN0 BPVI1/TDN1 BPVI2/TDN2 BPVI3/TDN3 BPVI4/TDN4 BPVI5/TDN5 BPVI6/TDN6 BPVI7/TDN7		38 31 79 72 109 102 7 144	N3 L3 L12 N12 B12 D12 D3 B3	on this pin. mode is as	Data on TDPn/T the follow:  TDPn 0 0 1 1 TDNn high for me	TDNn 0 1 0 1 ore than 16 co	Output Pulse Space Negative Pulse Positive Pulse Space	es of TCLKn. The line code in dual rail
TCLK0		36		TCLKn: Transmit Clock for Channel 0~7 The clock of 2.048 MHz for transmit is input on this pin. The transmit data at TDn/TDPn or TDNn is sampled into the device on the falling edges of TCLKn. Pulling TCLKn high for more than 16 MCLK cycles, the corresponding transmitter is set in Transmit All Ones (TAOS) state (when MCLK is clocked). In TAOS state, the TAOS generator adopts MCLK as the clock reference.  If TCLKn is low, the corresponding transmit channel is set into power down state, while driver output por become high-Z.  Different combinations of TCLKn and MCLK result in different transmit mode. It is summarized as the follows:				
TCLK1		29	N1 L1	MCLK	TCLKn		Tra	Insmit Mode
TCLK2		81	L14	Clocked	Clocked	Normal	peration	
TCLK3 TCLK4	I	74 107	N14 B14	Clocked	High (≥ 16 MC		t All Ones (TAOS) sigr channel.	nals to the line side in the corresponding
TCLK5		100	D14	Clocked	Low (≥ 64 MC			nannel is set into power down state.
TCLK6 TCLK7	9 2	D1 B1	High/Low	TCLK1 is cloc	TCLKn i: (≥ 16 TC TCLKn i: (≥ 64 TC The rece is high, a	CLK1) in the correspond CLK1) down state give path is not affected all receive paths just s	Il Ones (TAOS) signals to the line side esponding transmit channel.  ding transmit channel is set into power	
			1	[ <del>                                    </del>	TCLK1 is unava		: -, pr	

Table-1 Pin Description (Continued)

Name	Type	Pin No.		Description		
Name	Туре	TQFP144 PBGA160		Description		
				RDn: Receive Data for Channel 0~7		
				In Single Rail mode, the received NRZ data is output on this pin. The data is decoded by AMI or HDB3 lin		
RD0/RDP0		40	P2	code rule.		
RD1/RDP1		33	M2			
RD2/RDP2		77	M13	CVn: Code Violation for Channel 0~7		
RD3/RDP3		70	P13	In Single Rail mode, the bipolar violation, code violation and excessive zeros will be reported by driving p		
RD4/RDP4		111	A13	CVn high for a full clock cycle. However, only bipolar violation is indicated when AMI decoder is selected		
RD5/RDP5		104	C13			
RD6/RDP6		5	C2	RDPn/RDNn: Positive/Negative Receive Data for Channel 0~7		
RD7/RDP7	0	142	A2	In Dual Rail Mode with clock recovery, these pins output the NRZ data. A high signal on RDPn indicates the receipt of a positive pulse on RTIPn/RRINGn while a high signal on RDNn indicates the receipt of a		
CV0/RDN0	High-Z	41	P3	negative pulse on RTIPn/RRINGn.		
CV1/RDN1		34	M3	The output data at RDn or RDPn/RDNn are clocked out on the falling edges of RCLK when the CLKE input		
CV2/RDN2		76	M12	is low, or are clocked out on the rising edges of RCLK when CLKE is high.		
CV3/RDN3		69	P12	In Dual Rail Mode without clock recovery, these pins output the raw RZ sliced data. In this data recovery		
CV4/RDN4		112	A12	mode, the active polarity of RDPn/RDNn is determined by pin CLKE. When pin CLKE is low, RDPn/RDN		
CV5/RDN5		105	C12	is active low. When pin CLKE is high, RDPn/RDNn is active high.		
CV6/RDN6		4	C3	In hardware mode, RDn or RDPn/RDNn will remain active during LOS. In host mode, these pins will either		
CV7/RDN7		141	A3	remain active or insert alarm indication signal (AIS) into the receive path, determined by bit AISE in regis		
				ter GCF.		
				RDn or RDPn/RDNn is set into high-Z when the corresponding receiver is powered down.		
RCLK0		39	P1	RCLKn: Receive Clock for Channel 0~7		
RCLK1		32	M1	In clock recovery mode, this pin outputs the recovered clock from signal received on RTIPn/RRINGn. The		
RCLK2	0	78	M14	received data are clocked out of the device on the rising edges of RCLKn if pin CLKE is high, or on fallir		
RCLK3	U	71	P14	edges of RCLKn if pin CLKE is low.		
RCLK4	High-Z	110	A14	In data recovery mode, RCLKn is the output of an internal exclusive OR (XOR) which is connected with		
RCLK5	iligii-Z	103	C14	RDPn and RDNn. The clock is recovered from the signal on RCLKn.		
RCLK6		6	C1	If Receiver n is powered down, the corresponding RCLKn is in high-Z.		
RCLK7		143	A1	In Receiver this powered down, the corresponding RCLNH is in high-z.		
				MCLK: Master Clock		
				This is an independent, free running reference clock. A clock of 2.048 MHz is supplied to this pin as the clock reference of the device for normal operation.		
				In receive path, when MCLK is high, the device slices the incoming bipolar line signal into RZ pulse (Dai		
MCLK	1	10	E1	Recovery mode). When MCLK is low, all the receivers are powered down, and the output pins RCLKn,		
				RDPn and RDNn are switched to high-Z.		
				In transmit path, the operation mode is decided by the combination of MCLK and TCLKn (see TCLKn pi		
				description for details).		
				NOTE: Wait state generation via RDY/ACK is not available if MCLK is not provided.		
LOS0		42	K4	LOSn: Loss of Signal Output for Channel 0~7		
LOS1		35	K3	A high level on this pin indicates the loss of signal when there is no transition over a specified period of		
LOS2		75	K12	time or no enough ones density in the received signal. The transition will return to low automatically who		
LOS3	0	68	K11	there is enough transitions over a specified period of time with a certain ones density in the received sig		
LOS4		113	E11	nal. The LOS assertion and desertion criteria are described in 2.3.4 Loss of Signal (LOS) Detection.		
LOS5		106	E12			
LOS6		3	E3			
LOS7		140	E4			

Table-1 Pin Description (Continued)

	_	Pin	No.	Description			
Name	Туре	TQFP144	PBGA160				
	L	l	I.	Hardware/Hos	st Control Interface		
				MODE2: Control Mo			
				The signal on this pir	n determines which control mode is selected to control the device:		
				MODE	Control Interface		
				Low	Hardware Mode		
				VDDIO	0/2 Serial Host Interface		
				High	Parallel Host Interface		
MODE2	(Pulled to VDDIO/2)	11	E2	Serial host Interface Parallel host Interface	ns include MODE[2:0], LP[7:0], CODE, CLKE, JAS and OE. pins include $\overline{CS}$ , SCLK, SDI, SDO and $\overline{INT}$ . the pins include $\overline{CS}$ , A[4:0], D[7:0], $\overline{WR/DS}$ , $\overline{RD}/R/\overline{W}$ , ALE/ $\overline{AS}$ , $\overline{INT}$ and RDY/ $\overline{ACK}$ . The tiple parallel host interface as follows (refer to MODE1 and MODE0 pin descriptions		
				MODE[2:0]	Host Interface		
				100	Non-multiplexed Motorola Mode Interface		
				101	Non-multiplexed Intel Mode Interface		
				110	Multiplexed Motorola Mode Interface		
				111	Multiplexed Intel Mode Interface		
MODE1	I	43	K2	In parallel host mode, the parallel interface operates with separate address bus and data bus when this pir is low, and operates with multiplexed address and data bus when this pin is high.  In serial host mode or hardware mode, this pin should be grounded.			
					e, the parallel host interface is configured for Motorola compatible hosts when this pin mpatible hosts when this pin is high.		
MODE0/CODE	I	88	H12		Rule Select mode, the HDB3 encoder/decoder is enabled when this pin is low, and AMI encoder/when this pin is high. The selections affect all the channels.		
				In serial host mode, t	this pin should be grounded.		
	ı	87	J11		ctive Low) n is asserted low by the host to enable host interface. A high to low transition must each read/write operation and the level must not return to high until the operation is		
CS/JAS	(Pulled to			JAS: Jitter Attenuat In hardware control n	tor Select mode, this pin globally determines the Jitter Attenuator position:		
	VDDIO/2)			JAS	Jitter Attenuator (JA) Configuration		
				Low	JA in transmit path		
				VDDIO/2	JA not used		
				High	JA in receive path		
	1		1				

Table-1 Pin Description (Continued)

Name	Туре	Pin	No.	Description
Name	туре	TQFP144	PBGA160	Description
SCLK/ALE/AS	ı	86	J12	SCLK: Shift Clock In serial host mode, the signal on this pin is the shift clock for the serial interface. Data on pin SDO is clocked out on falling edges of SCLK if pin CLKE is high, or on rising edges of SCLK if pin CLKE is low. Data on pin SDI is always sampled on rising edges of SCLK.  ALE: Address Latch Enable In parallel Intel multiplexed host mode, the address on AD[4:0] is sampled into the device on the falling edges of ALE (signals on AD[7:5] are ignored). In non-multiplexed host mode, ALE should be pulled high.  AS: Address Strobe (Active Low)
				In parallel Motorola multiplexed host mode, the address on AD[4:0] is latched into the device on the falling edges of $\overline{AS}$ (signals on AD[7:5] are ignored). In non-multiplexed host mode, $\overline{AS}$ should be pulled high. <b>NOTE</b> : This pin is ignored in hardware control mode.
				RD: Read Strobe (Active Low) In parallel Intel multiplexed or non-multiplexed host mode, this pin is active low for read operation.
RD/R/W	I	85	J13	R/W: Read/Write Select In parallel Motorola multiplexed or non-multiplexed host mode, the pin is active low for write operation and high for read operation.  NOTE: This pin is ignored in hardware control mode.
SDI/WR/DS	I	84	J14	SDI: Serial Data Input In serial host mode, this pin input the data to the serial interface. Data on this pin is sampled on the rising edges of SCLK.  WR: Write Strobe (Active Low) In parallel Intel host mode, this pin is active low during write operation. The data on D[7:0] (in non-multiplexed mode) or AD[7:0] (in multiplexed mode) is sampled into the device on the rising edges of WR.  DS: Data Strobe (Active Low) In parallel Motorola host mode, this pin is active low. During a write operation ( $R/\overline{W} = 0$ ), the data on D[7:0] (in non-multiplexed mode) or AD[7:0] (in multiplexed mode) is sampled into the device on the rising edges of DS. During a read operation ( $R/\overline{W} = 1$ ), the data is driven to D[7:0] (in non-multiplexed mode) or AD[7:0] (in multiplexed mode) by the device on the rising edges of DS.  In parallel Motorola non-multiplexed host mode, the address information on the 5 bits of address bus A[4:0] are latched into the device on the falling edges of DS.  NOTE: This pin is ignored in hardware control mode.
SDO/RDY/ACK	0	83	K14	SDO: Serial Data Output In serial host mode, the data is output on this pin. In serial write operation, SDO is in high impedance for the first 8 SCLK clock cycles and driven low for the remaining 8 SCLK clock cycles. In serial read operation, SDO is in high-Z only when SDI is in address/command byte. Data on pin SDO is clocked out of the device on the falling edges of SCLK if pin CLKE is high, or on the rising edges of SCLK if pin CLKE is low.  RDY: Ready Output In parallel Intel host mode, the high level of this pin reports to the host that bus cycle can be completed, while low reports the host must insert wait states.  ACK: Acknowledge Output (Active Low) In parallel Motorola host mode, the low level of this pin indicates that valid information on the data bus is ready for a read operation or acknowledges the acceptance of the written data during a write operation.
ĪNT	O Open Drain	82	K13	INT: Interrupt (Active Low) This is an open drain, active low interrupt output. Three sources may cause the interrupt . Refer to 2.18 Interrupt Handling for details.

Table-1 Pin Description (Continued)

Nama	Tuna	Pin	No.	December			
Name	Туре	TQFP144	PBGA160	Description			
LP7/D7/AD7 LP6/D6/AD6 LP5/D5/AD5 LP4/D4/AD4 LP3/D3/AD3 LP2/D2/AD2 LP1/D1/AD1 LP0/D0/AD0	I/O High-Z	28 27 26 25 24 23 22 21	K1 J1 J2 J3 J4 H2 H3 G2	In hardware control mode, pin LPn configures the corresponding channel in different loopback mode, as follows:    LPn			
A4 MC3/A3 MC2/A2 MC1/A1 MC0/A0	I	12 13 14 15 16	F4 F3 F2 F1 G3				
OE	I	114	E14	mode, the signal on this pin is the address bus of the host interface.  OE: Output Driver Enable  Pulling this pin low can drive all driver output into high-Z for redundancy application without external mechanical relays. In this condition, all other internal circuits remain active.			

Table-1 Pin Description (Continued)

Name	Туре	Pin No.		Description	
Humo	1,700	TQFP144	PBGA160	Bookipaon	
CLKE	1	115	E13	CLKE: Clock Edge Select The signal on this pin determines the active edge of RCLKn and SCLK in clock recovery mode, or determines the active level of RDPn and RDNn in the data recovery mode. See 2.2 Clock Edges on page 14 ft details.	
		•		JTAG Signals	
	ı			TRST: JTAG Test Port Reset (Active Low)	
TRST	Pull-up	95	G12	This is the active low asynchronous reset to the JTAG Test Port. This pin has an internal pull-up resi and it can be left open.	
TMS	l Pull-up	96	F11	TMS: JTAG Test Mode Select  The signal on this pin controls the JTAG test performance and is clocked into the device on the rising edges of TCK. This pin has an internal pull-up resistor and it can be left open.	
тск	I	97	F14	TCK: JTAG Test Clock This pin input the clock of the JTAG Test. The data on TDI and TMS are clocked into the device on the ring edges of TCK, while the data on TDO is clocked out of the device on the falling edges of TCK. This p should be connected to GNDIO or VDDIO pin when unused.	
TDO	O High-Z	98	F13	TDO: JTAG Test Data Output This pin output the serial data of the JTAG Test. The data on TDO is clocked out of the device on a ing edges of TCK. TDO is a high-Z output signal. It is active only when scanning of data is out. This should be left float when unused.	
TDI	l Pull-up	99	F12	TDI: JTAG Test Data Input This pin input the serial data of the JTAG Test. The data on TDI is clocked into the device on the risin edges of TCK. This pin has an internal pull-up resistor and it can be left open.	
		1		Power Supplies and Grounds	
VDDIO	-	17 92	G1 G14	3.3 V I/O Power Supply	
GNDIO	-	18 91	G4 G11	I/O GND	
VDDT0		44	N4, P4		
VDDT1		53	L4, M4		
VDDT2		56 65	L11, M11	3.3 V/5 V Power Supply for Transmitter Driver	
VDDT3 VDDT4	-	65 116	N11, P11 A11, B11	All VDDT pins must be connected to 3.3 V or all VDDT must be connected to 5 V. It is not allowed to lea	
VDDT4 VDDT5		125	C11, D11	any of the VDDT pins open (not-connected) even if the channel is not used.	
VDDT6		128	C4, D4		
VDDT7		137	A4, B4		
GNDT0	+	47	N6, P6		
GNDT1		50	L6, M6		
GNDT2		59	L9, M9		
GNDT3			N9, P9	Analog GND for Transmitter Driver	
GNDT4	1 .	119	A9, B9	Alialog Grap for Transmitter priver	
GNDT5		122	C9, D9		
GNDT6		131	C6, D6		
GNDT7		134	A6, B6	36	
VDDD	-	19	H1	3.3 V Digital/Analog Core Power Supply	
VDDA	_	90	H14	old t Digitally fidicial office outphy	

Table-1 Pin Description (Continued)

Name	Pin No.		No.	Description	
Name	Type	TQFP144	PBGA160	Description	
GNDD GNDA	-	20 89	H4 H11	Digital/Analog Core GND	
	Others				
IC	0	93 94	G13 H13	IC: Internal Connection Internal use. Leave it float for normal operation.	

# 2 FUNCTIONAL DESCRIPTION

#### 2.1 OVERVIEW

The IDT82V2058 is a fully integrated octal short-haul line interface unit, which contains eight transmit and receive channels for use in E1 applications. The receiver performs clock and data recovery. As an option, the raw sliced data (no retiming) can be output to the system. Transmit equalization is implemented with low-impedance output drivers that provide shaped waveforms to the transformer, guaranteeing template conformance. A selectable jitter attenuator may be placed in the receive path or the transmit path. Moreover, multiple testing functions, such as error detection, loopback and JTAG boundary scan are also provided. The device is optimized for flexible software control through a serial or parallel host mode interface. Hardware control is also available. Figure-1 on page 1 shows one of the eight identical channels operation.

#### 2.1.1 SYSTEM INTERFACE

The system interface of each channel can be configured to operate in different modes:

- 1. Single rail interface with clock recovery.
- 2. Dual rail interface with clock recovery.
- 3. Dual rail interface with data recovery (that is, with raw data slicing only and without clock recovery).

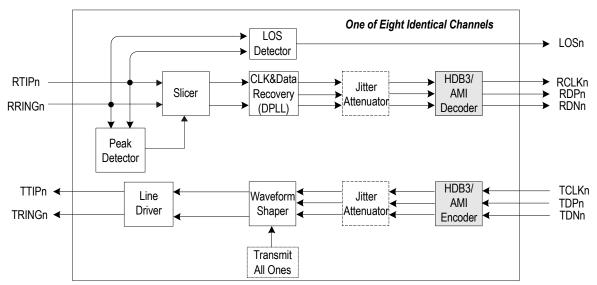
Each signal pin on system side has multiple functions depending on which operation mode the device is in.

The Dual Rail interface consists of TDPn<sup>1</sup>, TDNn, TCLKn, RDPn, RDNn and RCLKn. Data transmitted from TDPn and TDNn appears on TTIPn and TRINGn at the line interface; data received from the RTIPn and RRINGn at the line interface are transferred to RDPn and RDNn while the recovered clock extracting from the received data stream outputs on RCLKn. In Dual Rail operation, the clock/data recovery mode is selectable. Dual Rail interface with clock recovery shown in Figure-4 is a default configuration mode. Dual Rail interface with data recovery is shown in Figure-5. Pin RDPn and RDNn, are raw RZ slice outputs and internally connected to an EXOR which is fed to the RCLKn output for external clock recovery applications.

In Single Rail mode, data transmitted from TDn appears on TTIPn and TRINGn at the line interface. Data received from the RTIPn and RRINGn at the line interface appears on RDn while the recovered clock extracting from the received data stream outputs on RCLKn. When the device is in single rail interface, the selectable AMI or HDB3 line encoder/decoder is available and any code violation in the received data will be indicated at the CVn pin. The Single Rail mode has 2 sub-modes: Single Rail Mode 1 and Single Rail Mode 2. Single Rail Mode 1, whose interface is composed of TDn, TCLKn, RDn, CVn and RCLKn, is realized by pulling pin TDNn high for more than 16 consecutive TCLK cycles. Single Rail Mode 2, whose interface is composed of TDn, TCLKn, RDn, CVn, RCLKn and BPVIn, is realized by setting bit CRS in register e-CRS<sup>2</sup> and bit SING in register e-SING. The difference between them is that, in the latter mode bipolar violation can be inserted via pin BPVIn if AMI line code is selected.

The configuration of the Hardware Mode System Interface is summarized in Table-2. The configuration of the Host Mode System Interface is summarized in Table-3.

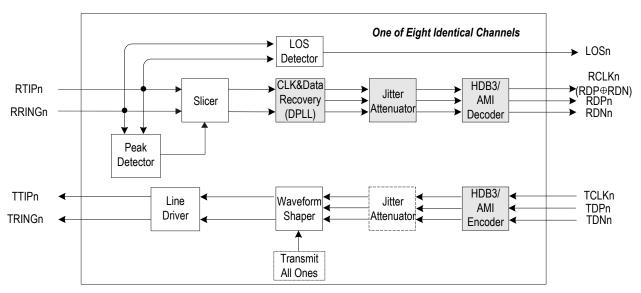
<sup>2.</sup> The first letter 'e-' indicates expanded register.



Note: The grey blocks are bypassed and the dotted blocks are selectable.

Figure-4 Dual Rail Interface with Clock Recovery

<sup>&</sup>lt;sup>1</sup> The footprint 'n' (n = 0 - 7) indicates one of the eight channels.



Note: The grey blocks are bypassed and the dotted blocks are selectable

Figure-5 Dual Rail Interface with Data Recovery

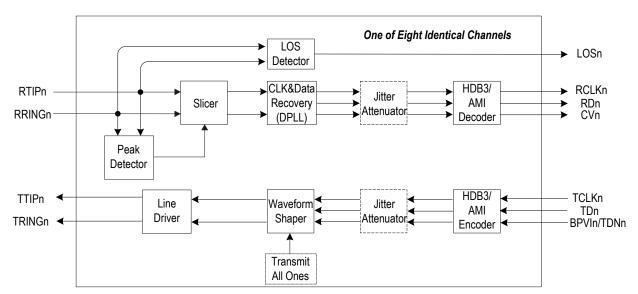


Figure-6 Single Rail Mode

Table-2 System Interface Configuration (In Hardware Mode)

Pin MCLK	Pin TDNn	Interface
Clocked	High (≥ 16 MCLK)	Single Rail Mode 1
Clocked	Pulse	Dual Rail mode with Clock Recovery
High	Pulse	Dual Rail mode with Data Recovery. Receive just slices the incoming data. Transmit is determined by the status of TCLKn.
Low	Pulse	Receiver is powered down. Transmit is determined by the status of TCLKn.

Table-3 System Interface Configuration (In Host Mode)

Pin MCLK	Pin TDNn	CRSn in e-CRS	SINGn in e-SING	Interface
Clocked	High	0	0	Single Rail Mode 1
Clocked	Pulse	0	1	Single Rail Mode 2
Clocked	Pulse	0	0	Dual Rail mode with Clock Recovery
Clocked	Pulse	1	0	Dual Rail mode with Data Recovery. Receive just slices the incoming data. Transmit is determined by the status of TCLKn.
High	Pulse	-	-	Dual Rail mode with Data Recovery. Receive just slices the incoming data. Transmit is determined by the status of TCLKn.
Low	Pulse	-	-	Receiver is powered down. Transmit is determined by the status of TCLKn.

Table-4 Active Clock Edge and Active Level

Pin CLKE	Pin RDn/F	RDPn and CVn/RDN	٧n	Pin SDO		
FIII CLKL	Clock Recove	ry	Slicer Output	FIII		
High	RCLKn	Active High	Active High	SCLK	Active High	
Low	RCLKn	Active High	Active Low	SCLK	Active High	

#### 2.2 CLOCK EDGES

The active edge of RCLKn and SCLK are selectable. If pin CLKE is high, the active edge of RCLKn is the rising edge, as for SCLK, that is falling edge. On the contrary, if CLKE is low, the active edge of RCLK is the falling edge and that of SCLK is rising edge. Pins RDn/RDPn, CVn/RDNn and SDO are always active high, and those output signals are clocked out on the active edge of RCLKn and SCLK respectively. See Table-4 Active Clock Edge and Active Level on page 14 for details. However, in dual rail mode without clock recovery, pin CLKE is used to set the active level for RDPn/RDNn raw slicing output: High for active high polarity and low for active low. It should be noted that data on pin SDI are always active high and are sampled on the rising edges of SCLK. The data on pin TDn/TDPn or BPVIn/TDNn are also always active high but are sampled on the falling edges of TCLKn, despite the level on CLKE.

#### 2.3 RECEIVER

In receive path, the line signals couple into RRINGn and RTIPn via a transformer and are converted into RZ digital pulses by a data slicer. Adaptation for attenuation is achieved using an integral peak detector that sets the slicing levels. Clock and data are recovered from the received RZ digital pulses by a digital phase-locked loop that provides jitter accommodation. After passing through the selectable jitter attenuator, the recovered data are decoded using HDB3 or AMI line code rules and clocked out of pin RDn in single rail mode, or presented on RDPn/RDNn in an undecoded dual rail NRZ format. Loss of signal, alarm indication signal, line code violation and excessive zeros are detected. These various changes in status may be enabled to generate interrupts.

#### 2.3.1 PEAK DETECTOR AND SLICER

The slicer determines the presence and polarity of the received pulses. In data recovery mode, the raw positive slicer output appears on RDPn while the negative slicer output appears on RDNn. In clock and

data recovery mode, the slicer output is sent to Clock and Data Recovery circuit for abstracting retimed data and optional decoding. The slicer circuit has a built-in peak detector from which the slicing threshold is derived. The slicing threshold is default to 50% (typical) of the peak value.

Signals with an attenuation of up to 12 dB (from 2.4 V) can be recovered by the receiver. To provide immunity from impulsive noise, the peak detectors are held above a minimum level of 0.150 V typically, despite the received signal level.

# 2.3.2 CLOCK AND DATA RECOVERY

The Clock and Data Recovery is accomplished by Digital Phase Locked Loop (DPLL). The DPLL is clocked 16 times of the received clock rate, i.e. 32.768 MHz in E1 mode. The recovered data and clock from DPLL is then sent to the selectable Jitter Attenuator or decoder for further processing.

The clock recovery and data recovery mode can be selected on a per channel basis by setting bit CRSn in register **e-CRS**. When bit CRSn is defaulted to '0', the corresponding channel operates in data and clock recovery mode. The recovered clock is output on pin RCLKn and retimed NRZ data are output on pin RDPn/RDNn in dual rail mode or on RDn in single rail mode. When bit CRSn is set to '1', dual rail mode with data recovery is enabled in the corresponding channel and the clock recovery is bypassed. In this condition, the analog line signals are converted to RZ digital bit streams on the RDPn/RDNn pins and internally connected to an EXOR which is fed to the RCLKn output for external clock recovery applications.

If MCLK is pulled high, all the receivers will enter the dual rail mode with data recovery. In this case, register **e-CRS** is ignored.

#### 2.3.3 HDB3/AMI LINE CODE RULE

Selectable HDB3 and AMI line coding/decoding is provided when the device is configured in single rail mode. HDB3 rules is enabled by setting bit CODE in register **GCF** to '0' or pulling pin CODE low. AMI rule is enabled by setting bit CODE in register **GCF** to '1' or pulling pin CODE high. The settings affect all eight channels.

Individual line code rule selection for each channel, if needed, is available by setting bit SINGn in register **e-SING** to '1' (to activate bit CODEn in register **e-CODE**) and programming bit CODEn to select line code rules in the corresponding channel: '0' for HDB3, while '1' for AMI. In this case, the value in bit CODE in register **GCF** or pin CODE for global control is unaffected in the corresponding channel and only affect in other channels.

In dual rail mode, the decoder/encoder are bypassed. Bit CODE in register **GCF**, bit CODEn in register **e-CODE** and pin CODE are ignored.

The configuration of the line code rule is summarized in Table-5.

# 2.3.4 LOSS OF SIGNAL (LOS) DETECTION

The Loss of Signal Detector monitors the amplitude and density of the received signal on receiver line before the transformer (measured on port A, B shown in Figure-10). The loss condition is reported by pulling pin LOSn high. At the same time, LOS alarm registers track LOS condition. When LOS is detected or cleared, an interrupt will generate if not masked. In host mode, the detection supports ITU G.775 and ETSI 300 233. In hardware mode, it supports the ITU G.775.

Table-6 summarizes the conditions of LOS in clock recovery mode.

During LOS, the RDPn/RDNn continue to output the sliced data when bit AISE in register **GCF** is set to '0' or output all ones as AIS (alarm indication signal) when bit AISE is set to '1'. The RCLKn is replaced by MCLK only if the bit AISE is set.

Table-5 Configuration of the Line Code Rule

	Hardware Mode					
CODE	Line Code Rule					
Low	All channels in HDB3					
High	All channels in AMI					

Host Mode									
CODE in GCF	Line Code Rule								
0	0/1	0	All channels in HDB3						
0	0	1	7 till Charlineis in Fibbo						
1	0/1	0	All channels in AMI						
1	1	1	All Charliners in Alvii						
0	1	1	CHn in AMI						
1	0	1	CHn in HDB3						

Table-6 LOS Condition in Clock Recovery Mode

		Standard					
		ETSI 300 233	LOSn				
LOS	Continuous Intervals	32	2048 (1 ms)	High			
Detected	Amplitude <sup>(1)</sup> below typical 200 mVp		below typical 200 mVp	l "igii			
LOS Cleared	Density	12.5% (4 marks in a sliding 32-bit period) with no more than 15 continuous zeros	12.5% (4 marks in a sliding 32-bit period) with no more than 15 continuous zeros	Low			
Cleared	Amplitude <sup>(1)</sup>	exceed typical 250 mVp	exceed typical 250 mVp				

<sup>1.</sup> LOS levels at device (RTIPn, RRINGn) with all ones signal. For more detail regarding the LOS parameters, please refer to Receiver Characteristics on page 42.

# 2.3.5 ALARM INDICATION SIGNAL (AIS) DETECTION

Alarm Indication Signal is available only in host mode with clock recovery, as shown in Table-7.

#### 2.3.6 ERROR DETECTION

The device can detect excessive zeros, bipolar violation and HDB3 code violation, as shown in Figure-7 and Figure-8. All the three kinds of errors are reported in both host mode and hardware mode with HDB3 line code rule used. In host mode, the **e-CZER** and **e-CODV** are used to

determine whether excessive zeros and code violation are reported respectively. When the device is configured in AMI decoding mode, only bipolar violation can be reported.

The error detection is available only in single rail mode in which the pin CVn/RDNn is used as error report output (CVn pin).

The configuration and report status of error detection are summarized in Table-8.

Table-7 AIS Condition

	ITU G.775 (Register LAC defaulted to '0')	ETSI 300 233 (Register LAC set to '1')			
AIS Detected	Less than 3 zeros contained in each of two consecutive 512-bit stream are received	Less than 3 zeros contained in a 512-bit stream are received			
AIS Cleared	3 or more zeros contained in each of two consecutive 512-bit stream are received	3 or more zeros contained in a 512-bit stream are received			

**Table-8 Error Detection** 

Hardware Mode						
Line Code	Pin CVn Reports					
AMI	Bipolar Violation					
HDB3	Bipolar Violation + Code Violation + Excessive Zeros					

Host Mode										
Line Code   CODVn in e-CODV   CZERn in e-CZER   Pin CVn Reports										
AMI	-	-	Bipolar Violation							
	0	0	Bipolar Violation + Code Violation							
HDB3	0	1	Bipolar Violation + Code Violation + Excessive Zeros							
סטטוו	1	0	Bipolar Violation							
	1	1	Bipolar Violation + Excessive Zeros							

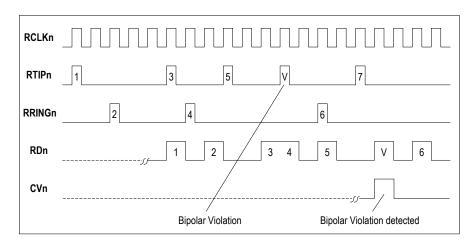


Figure-7 AMI Bipolar Violation

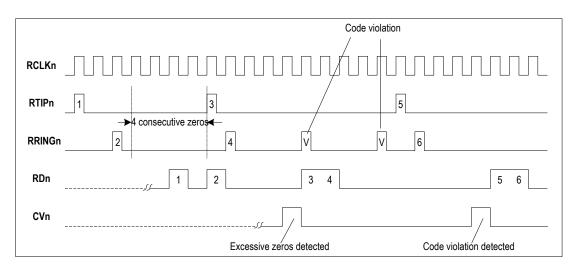


Figure-8 HDB3 Code Violation & Excessive Zeros

# 2.4 TRANSMITTER

In transmit path, data in NRZ format are clocked into the device on TDn and encoded by AMI or HDB3 line code rules when single rail mode is configured or pre-encoded data in NRZ format are input on TDPn and TDNn when dual rail mode is configured. The data are sampled into the device on falling edges of TCLKn. Jitter attenuator, if enabled, is

provided with a FIFO through which the data to be transmitted are passing. A low jitter clock is generated by an integral digital phase-locked loop and is used to read data from the FIFO. The shape of the pulses should meet the E1 pulse template after the signal passes through different cable lengths or types. Bipolar violation, for diagnosis, can be inserted on pin BPVIn if AMI line code rule is enabled.

#### 2.4.1 WAVEFORM SHAPER

E1 pulse template, specified in ITU-T G.703, is shown in Figure-9. The device has built-in transmit waveform templates for cable of 75  $\Omega$  or 120  $\Omega$ .

The built-in waveform shaper uses an internal high frequency clock which is 16XMCLK as the clock reference. This function will be bypassed when MCLK is unavailable.

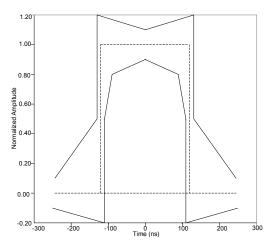


Figure-9 CEPT Waveform Template

#### 2.4.2 BIPOLAR VIOLATION INSERTION

When configured in Single Rail Mode 2 with AMI line code enabled, pin TDNn/BPVIn is used as BPVI input. A low-to-high transition on this pin inserts a bipolar violation on the next available mark in the transmit data stream. Sampling occurs on the falling edges of TCLK. But in TAOS (Transmit All Ones) with Analog Loopback and Remote Loopback, the BPVI is disabled. In TAOS with Digital Loopback, the BPVI is looped back to the system side, so the data to be transmitted on TTIPn and TRINGn are all ones with no bipolar violation.

#### 2.5 JITTER ATTENUATOR

The jitter attenuator can be selected to work either in transmit path or in receive path or not used. The selection is accomplished by setting pin JAS in hardware mode or configuring bits JACF[1:0] in register **GCF** in host mode, which affects all eight channels.

For applications which require line synchronization, the line clock needed to be extracted for the internal synchronization, the jitter attenuator is set in the receive path. Another use of the jitter attenuator is to provide clock smoothing in the transmit path for applications such as synchronous/asynchronous demultiplexing applications. In these applications, TCLK will have an instantaneous frequency that is higher than the nominal E1 data rate and in order to set the average long-term TCLK frequency within the transmit line rate specifications, periods of TCLK are suppressed (gapped).

The jitter attenuator integrates a FIFO which can accommodate a gapped TCLK. In host mode, the FIFO length can be 32 X 2 or 64 X 2 bits by programming bit JADP in **GCF**. In hardware mode, it is fixed to 64 X 2 bits. The FIFO length determines the maximum permissible gap width (see Table-9 Gap Width Limitation). Exceeding these values will cause FIFO overflow or underflow. The data is 16 or 32 bits' delay through the jitter attenuator in the corresponding transmit or receive path. The constant delay feature is crucial for the applications requiring "hitless" switching.

Table-9 Gap Width Limitation

FIFO Length	Max. Gap Width
64 bit	56 UI
32 bit	28 UI

In host mode, bit JABW in GCF determines the jitter attenuator 3 dB corner frequency (fc). In hardware mode, the fc is fixed to 1.7 Hz. Generally, the lower the fc is, the higher the attenuation. However, lower fc comes at the expense of increased acquisition time. Therefore, the optimum fc is to optimize both the attenuation and the acquisition time. In addition, the longer FIFO length results in an increased throughput delay and also influences the 3 dB corner frequency. Generally, it's recommended to use the lower corner frequency and the shortest FIFO length that can still meet jitter attenuation requirements.

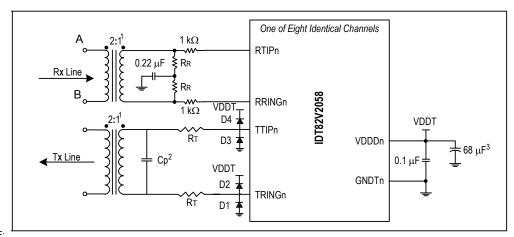
The output jitter meets ITU-T G.736, ITU-T G.742, ITU-T G.783 and ETSI CTR 12/13.

#### 2.6 LINE INTERFACE CIRCUITRY

The transmit and receive interface RTIPn/RRINGn and TTIPn/TRINGn connections provide a matched interface to the cable. Figure-10 shows the appropriate external components to connect with the cable for one transmit/receive channel. Table-10 summarizes the component values based on the specific application.

**Table-10 External Components Values** 

Component	<b>75</b> Ω <b>Coax</b>	120 Ω Twisted Pair				
$R_{\scriptscriptstyle T}$	$9.5~\Omega\pm1\%$	$9.5~\Omega\pm1\%$				
$R_R$	$9.31~\Omega\pm1\%$	15 $\Omega$ $\pm$ 1%				
Ср	2200 pF					
D1 - D4	Nihon Inter Electronics - EP05 E0 Motorola - MBR0540T1	Q03L, 11EQS03L, C10QS04, EC10QS03L;				



NOTE:

- 1. Pulse T1124 transformer is recommended to be used in Standard (STD) operating temperature range (0°C to 70°C), while Pulse T1114 transformer is recommended to be used in Extended (EXT) operating temperature range is -40°C to +85°C. See Transformer Specifications Table for details
- 2. Typical value. Adjust for actual board parasitics to obtain optimum return loss.
- 3. Common decoupling capacitor for all VDDT and GNDT pins. One per chip.

Figure-10 External Transmit/Receive Line Circuitry

## 2.7 TRANSMIT DRIVER POWER SUPPLY

All transmit driver power supplies must be 5.0 V or 3.3 V.

Despite the power supply voltage, the 75  $\Omega$ /120  $\Omega$  lines are driven through a pair of 9.5  $\Omega$  series resistors and a 1:2 transformer.

Table-11 Transformer Specifications<sup>(1)</sup>

	Electrical Specification @ 25°C										
Part	: No.	Turns Ratio (F	Pri: sec ± 2%)	ri: sec ± 2%) OCL @ 25°C (mH MIN)		L <sub>L</sub> (μΗ MAX)		C <sub>W/W</sub> (pF MAX)		Package/Schematic	
STD Temp.	EXT Temp.	Transmit	Receive	Transmit	Receive	Transmit	Receive	Transmit	Receive	1 dokago/odnomatio	
T1124	T1114	1:2CT	1CT:2	1.2	1.2	.6	.6	35	35	TOU/3	

<sup>1.</sup> Pulse T1124 transformer is recommended to be used in Standard (STD) operating temperature range (0°C to 70°C), while Pulse T1114 transformer is recommended to be used in Extended (EXT) operating temperature range is -40°C to +85°C.

#### 2.8 POWER DRIVER FAILURE MONITOR

An internal power Driver Failure Monitor (DFMON), parallel connected with TTIPn and TRINGn, can detect short circuit failure between TTIPn and TRINGn pins. Bit SCPB in register **GCF** decides whether the output driver short circuit protection is enabled. When the short circuit protection is enabled, the driver output current is limited to a typical value: 180 mAp. Also, register **DF**, **DFI** and **DFM** will be available. When DFMON will detect a short circuit, register **DF** will be set. With a short circuit failure detected and short circuit protection enabled, register **DFI** will be set and an interrupt will be generated on pin  $\overline{\text{INT}}$ .

# 2.9 TRANSMIT LINE SIDE SHORT CIRCUIT FAILURE DETECTION

A pair of  $9.5~\Omega$  serial resistors connect with TTIPn and TRINGn pins and limit the output current. In this case, the output current is a limited value which is always lower than the typical line short circuit current 180 mAp, even if the transmit line side is shorted.

Refer to Table-10 External Components Values for details.

#### 2.10 LINE PROTECTION

In transmit side, the Schottky diodes D1~D4 are required to protect the line driver and improve the design robustness. In receive side, the series resistors of 1 k $\Omega$  are used to protect the receiver against current surges coupled in the device. The series resistors do not affect the receiver sensitivity, since the receiver impedance is as high as 120 k $\Omega$  typically.

# 2.11 HITLESS PROTECTION SWITCHING (HPS)

The IDT82V2058 transceivers include an output driver with high-Z feature for E1 redundancy applications. This feature reduces the cost of redundancy protection by eliminating external relays. Details of HPS are described in relative Application Note.

#### 2.12 SOFTWARE RESET

Writing register **RS** will cause software reset by initiating about 1  $\mu s$  reset cycle. This operation set all the registers to their default value.

#### 2.13 POWER ON RESET

During power up, an internal reset signal sets all the registers to default values. The power-on reset takes at least 10  $\mu s$ , starting from when the power supply exceeds 2/3 VDDA.

#### 2.14 POWER DOWN

Each transmit channel will be powered down by pulling pin TCLKn low for more than 64 MCLK cycles (if MCLK is available) or about 30  $\mu$ s (if MCLK is not available). In host mode, each transmit channel will also be powered down by setting bit TPDNn in register **e-TPDN** to '1'.

All the receivers will be powered down when MCLK is low. When MCLK is clocked or high, setting bit RPDNn in register **e-RPDN** to '1' will configure the corresponding receiver to be powered down.

#### 2.15 INTERFACE WITH 5 V LOGIC

The IDT82V2058 can interface directly with 5 V TTL family devices. The internal input pads are tolerant to 5 V output from TTL and CMOS family devices.

#### 2.16 LOOPBACK MODE

The device provides four different diagnostic loopback configurations: Digital Loopback, Analog Loopback, Remote Loopback and Dual Loopback. In host mode, these functions are implemented by programming the registers **DLB**, **ALB** and **RLB** respectively. In hardware mode, only Analog Loopback and Remote Loopback can be selected by pin LPn.

#### 2.16.1 DIGITAL LOOPBACK

By programming the bits of register **DLB**, each channel of the device can be configured in Local Digital Loopback. In this configuration, the data and clock to be transmitted, after passing the encoder, are looped back to Jitter Attenuator (if enabled) and decoder in the receive path, then output on RCLKn, RDn/RDPn and CVn/RDNn. The data to be transmitted are still output on TTIPn and TRINGn while the data received on RTIPn and RRINGn are ignored. The Loss Detector is still in use. Figure-11 shows the process.

During Digital Loopback, the received signal on the receive line is still monitored by the LOS Detector (See 2.3.4 Loss of Signal (LOS) Detection for details). In case of a LOS condition and AIS insertion enabled, all ones signal will be output on RDPn/RDNn. With ATAO enabled, all ones signal will be also output on TTIPn/TRINGn. AIS insertion can be enabled by setting AISE bit in register **GCF** and ATAO can be enabled by setting register **ATAO** (default disabled).

#### 2.16.2 ANALOG LOOPBACK

By programming the bits of register ALB or pulling pin LPn high, each channel of the device can be configured in Analog Loopback. In this configuration, the data to be transmitted output from the line driver are internally looped back to the slicer and peak detector in the receive path and output on RCLKn, RDn/RDPn and CVn/RDNn. The data to be transmitted are still output on TTIPn and TRINGn while the data received on RTIPn and RRINGn are ignored. The LOS Detector (See 2.3.4 Loss of Signal (LOS) Detection for details) is still in use and monitors the internal looped back data. If a LOS condition on TDPn/TDNn is expected during Analog Loopback, ATAO should be disabled (default). Figure-12 shows the process.

The TTIPn and RTIPn, TRINGn and RRINGn cannot be connected directly to do the external analog loopback test. Line impedance loading is required to conduct the external analog loopback test.

#### 2.16.3 REMOTE LOOPBACK

By programming the bits of register **RLB** or pulling pin LPn low, each channel of the device can be set in Remote Loopback. In this configuration, the data and clock recovered by the clock and data recovery circuits are looped to waveform shaper and output on TTIPn and TRINGn. The jitter attenuator is also included in loopback when enabled in the transmit or receive path. The received data and clock are still output on RCLKn, RDn/RDPn and CVn/RDNn while the data to be transmitted on TCLKn, TDn/TDPn and BPVIn/TDNn are ignored. The LOs Detector is still in use. Figure-13 shows the process.

#### 2.16.4 DUAL LOOPBACK

Dual Loopback mode is set by setting bit DLBn in register **DLB** and bit RLBn in register **RLB** to '1'. In this configuration, after passing the encoder, the data and clock to be transmitted are looped back to decoder directly and output on RCLKn, RDn/RDPn and CVn/RDNn. The recovered data from RTIPn and RRINGn are looped back to waveform shaper through JA (if selected) and output on TTIPn and TRINGn. The LOS Detector is still in use. Figure-14 shows the process.

#### 2.16.5 TRANSMIT ALL ONES (TAOS)

In hardware mode, the TAOS mode is set by pulling pin TCLKn high for more than 16 MCLK cycles. In host mode, TAOS mode is set by programming register **TAO**. In addition, automatic TAOS signals are inserted by setting register **ATAO** when Loss of Signal occurs. Note that the TAOS generator adopts MCLK as a timing reference. In order to assure that the output frequency is within specified limits, MCLK must have the applicable stability.

The TAOS mode, the TAOS mode with Digital Loopback and the TAOS mode with Analog Loopback are shown in Figure-15, Figure-16 and Figure-17.

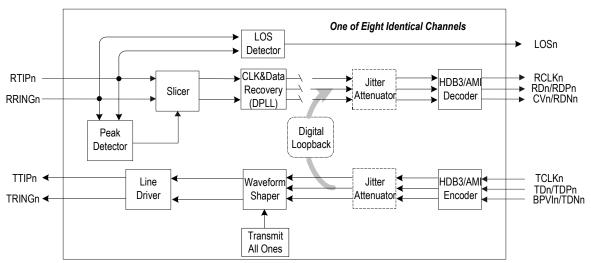


Figure-11 Digital Loopback

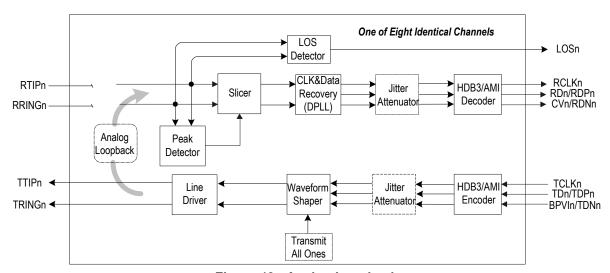


Figure-12 Analog Loopback

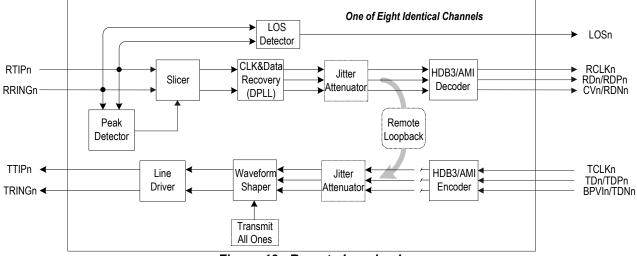


Figure-13 Remote Loopback

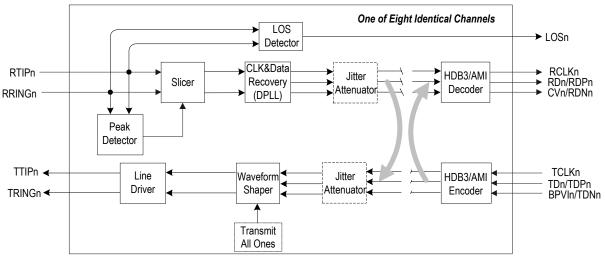


Figure-14 Dual Loopback

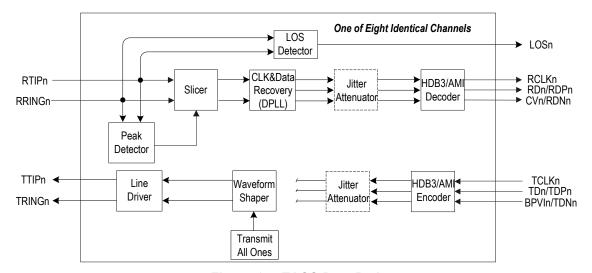


Figure-15 TAOS Data Path

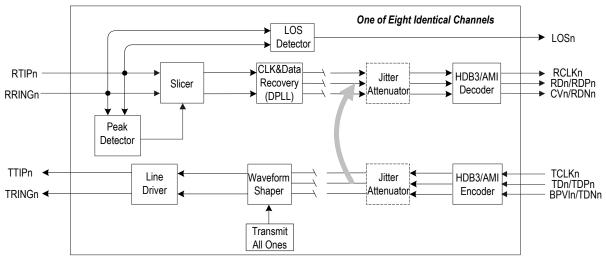


Figure-16 TAOS with Digital Loopback

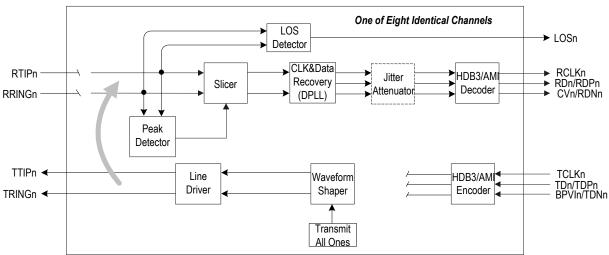


Figure-17 TAOS with Analog Loopback

#### 2.17 HOST INTERFACE

The host interface provides access to read and write the registers in the device. The interface consists of serial host interface and parallel host interface. By pulling pin MODE2 to VDDIO/2 or high, the device can be set to work in serial mode and in parallel mode respectively.

#### 2.17.1 PARALLEL HOST INTERFACE

The interface is compatible with Motorola and Intel host. Pins MODE1 and MODE0 are used to select the operating mode of the parallel host interface. When pin MODE1 is pulled low, the host uses separate address bus and data bus. When high, multiplexed address/data bus is used. When pin MODE0 is pulled low, the parallel host interface is configured for Motorola compatible hosts. When pin MODE0 is pulled high, the parallel host interface is configured for Intel compatible hosts. See Table-1 Pin Description for more details. The host interface pins in each operation mode is tabulated in Table-12:

Table-12 Parallel Host Interface Pins

MODE[2:0]	Host Interface	Generic Control, Data and Output Pin
100	Non-multiplexed Motorola interface	$\overline{\text{CS}}$ , $\overline{\text{ACK}}$ , $\overline{\text{DS}}$ , $R/\overline{W}$ , $\overline{\text{AS}}$ , A[4:0], D[7:0], $\overline{\text{INT}}$
101	Non-multiplexed Intel interface	CS, RDY, WR, RD, ALE, A[4:0], D[7:0], INT
110	Multiplexed Motorola interface	$\overline{\text{CS}}$ , $\overline{\text{ACK}}$ , $\overline{\text{DS}}$ , $\overline{\text{R/W}}$ , $\overline{\text{AS}}$ , $\overline{\text{AD[7:0]}}$ , $\overline{\text{INT}}$
111	Multiplexed Intel interface	CS. RDY. WR. RD. ALE. ADI7:01. INT

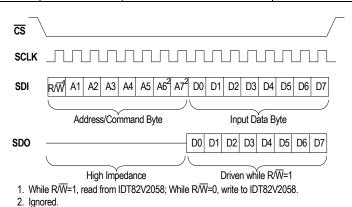


Figure-18 Serial Host Mode Timing

# 2.17.2 SERIAL HOST INTERFACE

By pulling pin MODE2 to VDDIO/2, the device operates in the serial host Mode. In this mode, the registers are accessible through a 16-bit word which contains an 8-bit command/address byte (bit  $R/\overline{W}$  and 5-address-bit A1~A5, A6 and A7 bits are ignored) and a subsequent 8-bit

data byte (D7~D0), as shown in Figure-18. When bit  $R/\overline{W}$  is set to '1', data is read out from pin SDO. When bit  $R/\overline{W}$  is set to '0', data on pin SDI is written into the register whose address is indicated by address bits A5~A1. See Figure-18 Serial Host Mode Timing.

## 2.18 INTERRUPT HANDLING

#### 2.18.1 INTERRUPT SOURCES

There are three kinds of interrupt sources:

- Status change in register LOS. The analog/digital loss of signal detector continuously monitors the received signal to update the specific bit in register LOS which indicates presence or absence of a LOS condition.
- Status change in register **DF**. The automatic power driver circuit continuously monitors the output drivers signal to update the specific bit in register **DFM** which indicates presence or absence of an output driver short circuit condition.
- Status change in register AIS. The AIS detector monitors the received signal to update the specific bit in register AIS which indicates presence or absence of a AIS condition.

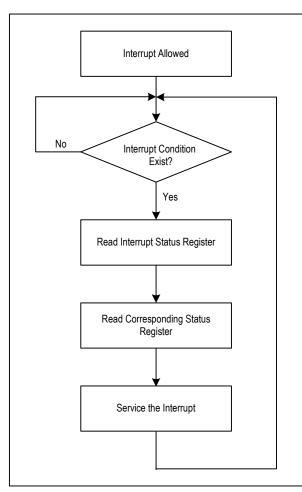


Figure-19 Interrupt Service Routine

#### 2.18.2 INTERRUPT ENABLE

The IDT82V2058 provides a latched interrupt output ( $\overline{\text{INT}}$ ) and the four kinds of interrupts are all reported by this pin. When the Interrupt Mask register (**LOSM**, **DFM** and **AISM**) is set to '1', the Interrupt Status register (**LOSI**, **DFI** and **AISI**) is enabled respectively. Whenever there is a transition ('0' to '1' or '1' to '0') in the corresponding status register, the Interrupt Status register will change into '1', which means an interrupt occurs, and there will be a high to low transition on  $\overline{\text{INT}}$  pin. An external pull-up resistor of approximately 10 k $\Omega$  is required to support the wire-OR operation of  $\overline{\text{INT}}$ . When any of the three Interrupt Mask registers is set to '0' (the power-on default value is '0'), the corresponding Interrupt Status register is disabled and the transition on status register is ignored.

#### 2.18.3 INTERRUPT CLEARING

When an interrupt occurs, the Interrupt Status registers: LOSI, DFI and AISI, are read to identify the interrupt source. These registers will be cleared to '0' after the corresponding status registers: LOS, DF and AIS are read. The Status registers will be cleared once the corresponding conditions are met.

Pin INT is pulled high when there is no pending interrupt left. The interrupt handling in the interrupt service routine is showed in Figure-19.

#### 2.19 **G.772 MONITORING**

The eight channels of IDT82V2058 can all be configured to work as regular transceivers. In applications using only seven channels (channels 1 to 7), channel 0 is configured to non-intrusively monitor any of the other channels' inputs or outputs on the line side. The monitoring is non-intrusive per ITU-T G.772. Figure-20 shows the Monitoring Principle. The receiver path or transmitter path to be monitored is configured by pins MC[3:0] in hardware mode or by register **PMON** in host mode.

The monitored signal goes through the clock and data recovery circuit of channel 0. The monitored clock can output on RCLK0 which can be used as a timing interfaces derived from E1 signal. The monitored data can be observed digitally at the output pins RCLK0, RD0/RDP0 and RDN0. LOS detector is still in use in channel 0 for the monitored signal.

In monitoring mode, channel 0 can be configured in Remote Loopback. The signal which is being monitored will output on TTIP0 and TRING0. The output signal can then be connected to a standard test equipment with an E1 electrical interface for non-intrusive monitoring.

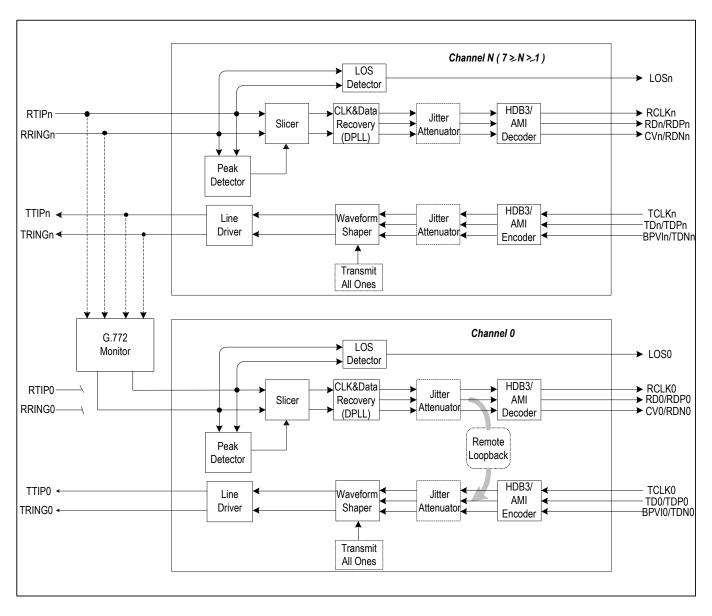


Figure-20 Monitoring Principle

# 3 PROGRAMMING INFORMATION

#### 3.1 REGISTER LIST AND MAP

There are 21 primary registers (including an Address Pointer Control Register and 8 expanded registers in the device).

Whatever the control interface is, 5 address bits are used to set the registers. In non-multiplexed parallel interface mode, the five dedicated address bits are A[4:0]. In multiplexed parallel interface mode, AD[4:0] carries the address information. In serial interface mode, A[5:1] are used to address the register.

The Register **ADDP**, addressed as 11111 or 1F Hex, switches between primary registers bank and expanded registers bank.

By setting the register **ADDP** to 'AAH', the 5 address bits point to the expanded register bank, that is, the expanded registers are available. By clearing register **ADDP**, the primary registers are available.

Primary Registers, whose addresses are 10H, 11H, 16H to 1EH, are reserved. Expanded registers, whose addresses are 08H to 1EH, are used for test and must be set to '0' (default).

**Table-13 Primary Register List** 

	Addres	SS	Register	R/W	Explanation				
Hex	Serial Interface A7-A1	Parallel Interface A7-A0	Register	IK/VV	Explanation				
00	XX00000	XXX00000	ID	R	Device ID Register				
01	XX00001	XXX00001	ALB	R/W	Analog Loopback Configuration Register				
02	XX00010	XXX00010	RLB	R/W	Remote Loopback Configuration Register				
03	XX00011	XXX00011	TAO	R/W	Transmit All Ones Configuration Register				
04	XX00100	XXX00100	LOS	R	Loss of Signal Status Register				
05	XX00101	XXX00101	DF	R	Driver Fault Status Register				
06	XX00110	XXX00110	LOSM	R/W	LOS Interrupt Mask Register				
07	XX00111	XXX00111	DFM	R/W	Driver Fault Interrupt Mask Register				
08	XX01000	XXX01000	LOSI	R	LOS Interrupt Status Register				
09	XX01001	XXX01001	DFI	R	Driver Fault Interrupt Status Register				
0A	XX01010	XXX01010	RS	W	Software Reset Register				
0B	XX01011	XXX01011	PMON	R/W	Performance Monitor Configuration Register				
0C	XX01100	XXX01100	DLB	R/W	Digital Loopback Configuration Register				
0D	XX01101	XXX01101	LAC	R/W	LOS/AIS Criteria Configuration Register				
0E	XX01110	XXX01110	ATAO	R/W	Automatic TAOS Configuration Register				
0F	XX01111	XXX01111	GCF	R/W	Global Configuration Register				
10	XX10000	XXX10000			Reserved				
11	XX10001	XXX10001							
12	XX10010	XXX10010	OE	R/W	Output Enable Configuration Register				
13	XX10011	XXX10011	AIS	R	AIS Status Register				
14	XX10100	XXX10100	AISM	R/W	AIS Interrupt Mask Register				
15	XX10101	XXX10101	AISI	R	AIS Interrupt Status Register				
16	XX10110	XXX10110							
17	XX10111	XXX10111							
18	XX11000	XXX11000							
19	XX11001	XXX11001							
1A	XX11010	XXX11010			Reserved				
1B	XX11011	XXX11011							
1C	XX11100	XXX11100							
1D	XX11101	XXX11101							
1E	XX11110	XXX11110							
1F	XX11111	XXX11111	ADDP	R/W	Address pointer control Register for switching between primary register bank and expanded register bank				

Table-14 Expanded (Indirect Address Mode) Register List

	Address		Address		Address		Register	R/W	Explanation
Hex	Serial Interface A7-A1	Parallel Interface A7-A0	3						
00	XX00000	XXX00000	e-SING	R/W	Single Rail Mode Setting Register				
01	XX00001	XXX00001	e-CODE	R/W					
02	XX00010	XXX00010	e-CRS	R/W					
03	XX00011	XXX00011	e-RPDN	R/W	Receiver n Powerdown Enable/Disable Register				
04	XX00100	XXX00100	e-TPDN	R/W	Transmitter n Powerdown Enable/Disable Register				
05	XX00101	XXX00101	e-CZER	R/W	Consecutive Zero Detect Enable/Disable Register				
06	XX00110	XXX00110	e-CODV	R/W	Code Violation Detect Enable/Disable Register				
07	XX00111	XXX00111	e-EQUA	R/W	Enable Equalizer Enable/Disable Register				
80	XX01000	XXX01000			<u> </u>				
09	XX01001	XXX01001							
0A	XX01010	XXX01010							
0B	XX01011	XXX01011							
0C	XX01100	XXX01100							
0D	XX01101	XXX01101							
0E	XX01110	XXX01110							
0F	XX01111	XXX01111							
10	XX10000	XXX10000							
11	XX10001	XXX10001							
12	XX10010	XXX10010							
13	XX10011	XXX10011			Test				
14	XX10100	XXX10100							
15	XX10101	XXX10101							
16	XX10110	XXX10110							
17	XX10111	XXX10111							
18	XX11000	XXX11000							
19	XX11001	XXX11001							
1A 1B	XX11010	XXX11010							
	XX11011	XXX11011							
1C 1D	XX11100 XX11101	XXX11100 XXX11101							
1E	XX11101 XX11110	XXX11101 XXX11110							
IE	ΛΛΙΙΙΙΟ	ΛΛΛΙΙΙΙ <b>υ</b>							
1F	XX11111	XXX11111	ADDP	R/W	Address pointer control register for switching between primary register bank and expanded register bank				

Table-15 Primary Register Map

Register	Address R/W Default	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
ID	00H	ID 7	ID 6	ID 5	ID 4	ID 3	ID 2	ID 1	ID 0
	R	R	R	R	R	R	R	R	R
	Default	0	0	0	1	0	0	0	0
ALB	01H	ALB 7	ALB 6	ALB 5	ALB 4	ALB 3	ALB 2	ALB 1	ALB 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
RLB	02H	RLB 7	RLB 6	RLB 5	RLB 4	RLB 3	RLB 2	RLB 1	RLB 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
TAO	03H	TAO 7	TAO 6	TAO 5	TAO 4	TAO 3	TAO 2	TAO 1	TAO 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
LOS	04H	LOS 7	LOS 6	LOS 5	LOS 4	LOS 3	LOS 2	LOS 1	LOS 0
	R	R	R	R	R	R	R	R	R
	Default	0	0	0	0	0	0	0	0
DF	05H	DF 7	DF 6	DF 5	DF 4	DF 3	DF 2	DF 1	DF 0
	R	R	R	R	R	R	R	R	R
	Default	0	0	0	0	0	0	0	0
LOSM	06H	LOSM 7	LOSM 6	LOSM 5	LOSM 4	LOSM 3	LOSM 2	LOSM 1	LOSM 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
DFM	07H	DFM 7	DFM 6	DFM 5	DFM 4	DFM 3	DFM 2	DFM 1	DFM 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
LOSI	08H	LOSI 7	LOSI 6	LOSI 5	LOSI 4	LOSI 3	LOSI 2	LOSI 1	LOSI 0
	R	R	R	R	R	R	R	R	R
	Default	0	0	0	0	0	0	0	0
DFI	09H	DFI 7	DFI 6	DFI 5	DFI 4	DFI 3	DFI 2	DFI 1	DFI 0
	R	R	R	R	R	R	R	R	R
	Default	0	0	0	0	0	0	0	0
RS	0AH	RS 7	RS 6	RS 5	RS 4	RS 3	RS 2	RS 1	RS 0
	W	W	W	W	W	W	W	W	W
	Default	1	1	1	1	1	1	1	1
PMON	0BH	-	-	-	-	MC 3	MC 2	MC 1	MC 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
DLB	0CH	DLB 7	DLB 6	DLB 5	DLB 4	DLB 3	DLB 2	DLB 1	DLB 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
LAC	0DH	LAC 7	LAC 6	LAC 5	LAC 4	LAC 3	LAC 2	LAC 1	LAC 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
ATAO	0EH	ATAO 7	ATAO 6	ATAO 5	ATAO 4	ATAO 3	ATAO 2	ATAO 1	ATAO 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
GCF	0FH	-	AISE	SCPB	CODE	JADP	JABW	JACF 1	JACF 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0

**Table-15 Primary Register Map (Continued)** 

Register	Address R/W Default	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
OE	12 Hex	OE 7	OE 6	OE 5	OE 4	OE 3	OE 2	OE 1	OE 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
AIS	13 Hex	AIS 7	AIS 6	AIS 5	AIS 4	AIS 3	AIS 2	AIS 1	AIS 0
	R	R	R	R	R	R	R	R	R
	Default	0	0	0	0	0	0	0	0
AISM	14 Hex	AISM 7	AISM 6	AISM 5	AISM 4	AISM 3	AISM 2	AISM 1	AISM 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
AISI	15 Hex	AISI 7	AISI 6	AISI 5	AISI 4	AISI 3	AISI 2	AISI 1	AISI 0
	R	R	R	R	R	R	R	R	R
	Default	0	0	0	0	0	0	0	0
ADDP	1F Hex	ADDP 7	ADDP 6	ADDP 5	ADDP 4	ADDP 3	ADDP 2	ADDP 1	ADDP 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0

Table-16 Expanded (Indirect Address Mode) Register Map

Register	Address R/W Default	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
e-SING	00H	SING 7	SING 6	SING 5	SING 4	SING 3	SING 2	SING 1	SING 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
e-CODE	01H R/W Default	CODE 7 R/W 0	CODE 6 R/W 0	CODE 5 R/W 0	CODE 4 R/W 0	CODE 3 R/W 0	CODE 2 R/W 0	CODE 1 R/W 0	CODE 0 R/W 0
e-CRS	02H	CRS 7	CRS 6	CRS 5	CRS 4	CRS 3	CRS 2	CRS 1	CRS 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
e-RPDN	03H	RPDN 7	RPDN 6	RPDN 5	RPDN 4	RPDN 3	RPDN 2	RPDN 1	RPDN 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
e-TPDN	04H	TPDN 7	TPDN 6	TPDN 5	TPDN 4	TPDN 3	TPDN 2	TPDN 1	TPDN 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
e-CZER	05H	CZER 7	CZER 6	CZER 5	CZER 4	CZER 3	CZER 2	CZER 1	CZER 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
e-CODV	06H	CODV 7	CODV 6	CODV 5	CODV 4	CODV 3	CODV 2	CODV 1	CODV 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
e-EQUA	07H	EQUA 7	EQUA 6	EQUA 5	EQUA 4	EQUA 3	EQUA 2	EQUA 1	EQUA 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0
ADDP	1FH	ADDP 7	ADDP 6	ADDP 5	ADDP 4	ADDP 3	ADDP 2	ADDP 1	ADDP 0
	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
	Default	0	0	0	0	0	0	0	0

#### 3.2 REGISTER DESCRIPTION

#### 3.2.1 PRIMARY REGISTERS

**ID:** Device ID Register (R, Address = 00H)

Symbol	Position	Default	Description
ID[7:0]	ID.7-0	10H	An 8-bit word is pre-set into the device as the identification and revision number. This number is different with the functional changes and is mask programmed.

# **ALB:** Analog Loopback Configuration Register (R/W, Address = 01H)

Symbol	Position	Default	Description
ALB[7:0]	ALB.7-0	00H	0 = Normal operation. (Default) 1 = Analog Loopback enabled.

# **RLB:** Remote Loopback Configuration Register (R/W, Address = 02H)

Symbol	Position	Default	Description
RLB[7:0]	RLB.7-0	00H	0 = Normal operation. (Default) 1 = Remote Loopback enabled.

# **TAO**: Transmit All Ones Configuration Register (R/W, Address = 03H)

Symbol	Position	Default	Description
TAO[7:0]	TAO.7-0	00H	0 = Normal operation. (Default) 1 = Transmit all ones.

# **LOS:** Loss of Signal Status Register (R, Address = 04H)

Symbol	Position	Default	Description
LOS[7:0]	LOS.7-0	00H	0 = Normal operation. (Default) 1 = Loss of signal detected.

# **DF:** Driver Fault Status Register (R, Address = 05H)

# **LOSM:** Loss of Signal Interrupt Mask Register (R/W, Address = 06H)

Symbol	Position	Default	Description
LOSM[7:0]	LOSM.7-0	00H	0 = LOS interrupt is not allowed. (Default) 1 = LOS interrupt is allowed.

# **DFM:** Driver Fault Interrupt Mask Register (R/W, Address = 07H)

Symbol	Position	Default	Description
DFM[7:0]	DFM.7-0	00H	0 = Driver fault interrupt not allowed. (Default) 1 = Driver fault interrupt allowed.

# **LOSI:** Loss of Signal Interrupt Status Register (R, Address = 08H)

Symbol	Position	Default	Description
LOSI[7:0]	LOSI.7-0	00H	0 = (Default). Or after a <b>LOS</b> read operation. 1 = Any transition on <b>LOSn</b> (Corresponding <b>LOSMn</b> is set to '1').

# **DFI:** Driver Fault Interrupt Status Register (R, Address = 09H)

Symbol	Position	Default	Description
DFI[7:0]	DFI.7-0	00H	0 = (Default). Or after a <b>DF</b> read operation. 1 = Any transition on <b>DFn</b> (Corresponding <b>DFMn</b> is set to '1').

# **RS:** Software Reset Register (W, Address = 0AH)

Symbol	Position	Default	Description
RS[7:0]	RS.7-0	FFH	Writing to this register will not change the content in this register but initiate a 1 µs reset cycle, which means all the registers in the device are set to their default values.

# **PMON:** Performance Monitor Configuration Register (R/W, Address = 0BH)

Symbol	Position	Default	Description
-	PMON.7-4	0000	0 = Normal operation. (Default) 1 = Reserved.
MC[3:0]	PMON.3-0	0000	0000 = Normal operation without monitoring (Default) 0001 = Monitor Receiver 1 0010 = Monitor Receiver 2 0011 = Monitor Receiver 3 0100 = Monitor Receiver 4 0101 = Monitor Receiver 5 0110 = Monitor Receiver 6 0111 = Monitor Receiver 7 1000 = Normal operation without monitoring 1001 = Monitor Transmitter 1 1010 = Monitor Transmitter 2 1011 = Monitor Transmitter 3 1100 = Monitor Transmitter 4 1101 = Monitor Transmitter 5 1110 = Monitor Transmitter 6 1111 = Monitor Transmitter 7

# **DLB**: Digital Loopback Configuration Register (R/W, Address = 0CH)

Symbol	Position	Default	Description
DLB[7:0]	DLB.7-0	00H	0 = Normal operation. (Default) 1 = Digital Loopback enabled.

# **LAC:** LOS/AIS Criteria Configuration Register (R/W, Address = 0DH)

Symbol	Position	Default	Description
LAC[7:0]	LAC.7-0	00H	0 = G.775 (Default) 1 = ETSI 300 233

# **ATAO**: Automatic TAOS Configuration Register (R/W, Address = 0EH)

Symbol	Position	Default	Description
ATAO[7:0]	ATAO.7-0	00H	0 = No automatic transmit all ones. (Default) 1 = Automatic transmit all ones to the line side during LOS.

# **GCF:** Global Configuration Register (R/W, Address = 0FH)

Symbol	Position	Default	Description
-	GCF.7	0	0 = Normal operation. 1 = Reserved.
AISE	GCF.6	0	0 = AIS insertion to the system side disabled on LOS. 1 = AIS insertion to the system side enabled on LOS.
SCPB	GCF.5	0	0 = Short circuit protection is enabled. 1 = Short circuit protection is disabled.
CODE	GCF.4	0	0 = HDB3 encoder/decoder enabled. 1 = AMI encoder/decoder enabled.
JADP	GCF.3	0	Jitter Attenuator Depth Select 0 = 32-bit FIFO (Default) 1 = 64-bit FIFO
JABW	GCF.2	0	Jitter Transfer Function Bandwidth Select 0 = 1.7 Hz 1 = 6.6 Hz
JACF[1:0]	GCF.1-0	00	Jitter Attenuator Configuration  00 = JA not used. (Default)  01 = JA in transmit path  10 = JA not used.  11 = JA in receive path

# **OE:** Output Enable Configuration Register (R/W, Address = 12H)

Symbol	Position	Default	Description
OE[7:0]	OE.7-0	00H	0 = Transmit drivers enabled. (Default) 1 = Transmit drivers in high-Z.

# **AIS:** Alarm Indication Signal Status Register (R, Address = 13H)

Symbol	Position	Default	Description
AIS[7:0]	AIS.7-0	00H	0 = Normal operation. (Default) 1 = AIS detected.

# AISM: Alarm Indication Signal Interrupt Mask Register (R/W, Address = 14H)

Symbol	Position	Default	Description
AISM[7:0]	AISM.7-0	00H	0 = AIS interrupt is not allowed. (Default) 1 = AIS interrupt is allowed.

# AISI: Alarm Indication Signal Interrupt Status Register (R, Address = 15H)

Symbol	Position	Default	Description
AISI[7:0]	AISI.7-0	00H	0 = (Default), or after an <b>AIS</b> read operation 1 = Any transition on <b>AISn</b> . (Corresponding <b>AISMn</b> is set to '1'.)

# **ADDP:** Address Pointer Control Register (R/W, Address = 1F H)

Symbol	Position	Default	Description
ADDP[7:0]	ADDP.7-0	00H	Two kinds of configuration in this register can be set to switch between primary register bank and expanded register bank.  When power up, the address pointer will point to the top address of primary register bank automatically.  00H = The address pointer points to the top address of primary register bank (default).  AAH = The address pointer points to the top address of expanded register bank.

#### 3.2.2 EXPANDED REGISTER DESCRIPTION

e-SING: Single Rail Mode Setting Register (R/W, Expanded Address = 00H)

Symbol	Position	Default	Description
SING[7:0]	SING.7-0	00H	0 = Pin TDNn selects single rail mode or dual rail mode. (Default) 1 = Single rail mode enabled (with CRSn=0)

#### **e-CODE**: Encoder/Decoder Selection Register (R/W, Expanded Address = 01H)

Symbol	Position	Default	Description
CODE[7:0]	CODE.7-0	00H	CODEn selects AMI or HDB3 encoder/decoder on a per channel basis with SINGn = 1 and CRSn = 0.  0 = HDB3 encoder/decoder enabled. (Default)  1 = AMI encoder/decoder enabled.

#### e-CRS: Clock Recovery Enable/Disable Selection Register (R/W, Expanded Address = 02H)

Symbol	Position	Default	Description
CRS[7:0]	CRS.7-0	00H	0 = Clock recovery enabled. (Default) 1 = Clock recovery disabled.

## e-RPDN: Receiver n Powerdown Register (R/W, Expanded Address = 03H)

Symbol	Position	Default	Description			
RPDN[7:0]	RPDN.7-0	00H	0 = Normal operation. (Default) 1 = Receiver n is powered down.			

# e-TPDN: Transmitter n Powerdown Register (R/W, Expanded Address = 04H)

Symbol	Position	Default	Description
TPDN[7:0]	TPDN.7-0	00H	0 = Normal operation. (Default) 1 = Transmitter n is powered down <sup>(1)</sup> (the corresponding transmit output driver enters a low power high-Z mode).

<sup>&</sup>lt;sup>1</sup> Transmitter n is powered down when either pin TCLKn is pulled low or TPDNn is set to '1'

# **e-CZER:** Consecutive Zero Detect Enable/Disable Register (R/W, Expanded Address = 05H)

Symbol	Position	Default	Description
CZER[7:0]	CZER.7-0	00H	0 = Excessive zeros detect disabled. (Default) 1 = Excessive zeros detect enabled for HDB3 decoder in single rail mode.

#### **e-CODV:** Code Violation Detect Enable/Disable Register (R/W, Expanded Address = 06H)

Symbol	Position	Default	Description
CODV[7:0]	CODV.7-0	00H	0 = Code Violation Detect enable for HDB3 decoder in single rail mode. (Default) 1 = Code Violation Detect disabled.

# e-EQUA: Receive Equalizer Enable/Disable Register (R/W, Expanded Address = 07H)

Symbol	Position	Default	Description
EQUA[7:0]	EQUA.7-0	00H	0 = Normal operation. (Default) 1 = Equalizer in Receiver n is enabled, which can improve the receive performance when transmission length is more than 200 m.

# 4 IEEE STD 1149.1 JTAG TEST ACCESS PORT

The IDT82V2058 supports the digital Boundary Scan Specification as described in the IEEE 1149.1 standards.

The boundary scan architecture consists of data and instruction registers plus a Test Access Port (TAP) controller. Control of the TAP is achieved through signals applied to the TMS and TCK pins. Data is shifted into the registers via the TDI pin, and shifted out of the registers via the TDO pin. JTAG test data are clocked at a rate determined by JTAG test clock.

The JTAG boundary scan registers includes BSR (Boundary Scan Register), IDR (Device Identification Register), BR (Bypass Register) and IR (Instruction Register). These will be described in the following pages. Refer to Figure-21 for architecture.

# 4.1 JTAG INSTRUCTIONS AND INSTRUCTION REGISTER (IR)

The IR with instruction decode block is used to select the test to be executed or the data register to be accessed or both.

The instructions are shifted in LSB first to this 3-bit register. See Table-17 Instruction Register Description on page 34 for details of the codes and the instructions related.

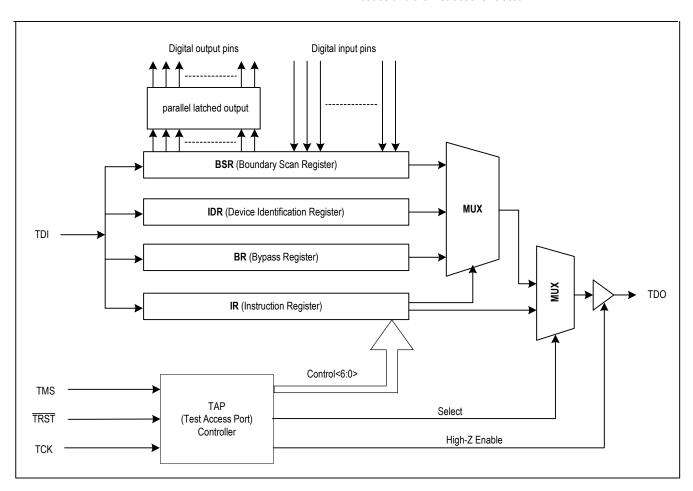


Figure-21 JTAG Architecture

**Table-17 Instruction Register Description** 

IR Code	Instruction	Comments
000	Extest	The external test instruction allows testing of the interconnection to other devices. When the current instruction is the EXTEST instruction, the boundary scan register is placed between TDI and TDO. The signal on the input pins can be sampled by loading the boundary scan register using the Capture-DR state. The sampled values can then be viewed by shifting the boundary scan register using the Shift-DR state. The signal on the output pins can be controlled by loading patterns shifted in through input TDI into the boundary scan register using the Update-DR state.
100	Sample/Preload	The sample instruction samples all the device inputs and outputs. For this instruction, the boundary scan register is placed between TDI and TDO. The normal path between IDT82V2058 logic and the I/O pins is maintained. Primary device inputs and outputs can be sampled by loading the boundary scan register using the Capture-DR state. The sampled values can then be viewed by shifting the boundary scan register using the Shift-DR state.
110	Idcode	The identification instruction is used to connect the identification register between TDI and TDO. The device's identification code can then be shifted out using the Shift-DR state.
111	Bypass	The bypass instruction shifts data from input TDI to output TDO with one TCK clock period delay. The instruction is used to bypass the device.

**Table-18 Device Identification Register Description** 

Bit No.	Comments
0	Set to '1'
1~11	Producer Number
12~27	Part Number
28~31	Device Revision

#### 4.2 JTAG DATA REGISTER

#### 4.2.1 DEVICE IDENTIFICATION REGISTER (IDR)

The IDR can be set to define the producer number, part number and the device revision, which can be used to verify the proper version or revision number that has been used in the system under test. The IDR is 32 bits long and is partitioned as in Table-18. Data from the IDR is shifted out to TDO LSB first.

#### **Table-19 Boundary Scan Register Description**

Bit No.	Bit Symbol	Pin Signal	Туре	Comments
0	POUT0	LP0	I/O	
1	PIN0	LP0	I/O	
2	POUT1	LP1	I/O	
3	PIN1	LP1	I/O	
4	POUT2	LP2	I/O	
5	PIN2	LP2	I/O	
6	POUT3	LP3	I/O	
7	PIN3	LP3	I/O	
8	POUT4	LP4	I/O	
9	PIN4	LP4	I/O	
10	POUT5	LP5	I/O	
11	PIN5	LP5	I/O	
12	POUT6	LP6	I/O	
13	PIN6	LP6	I/O	
14	POUT7	LP7	I/O	
15	PIN7	LP7	I/O	

#### 4.2.2 BYPASS REGISTER (BR)

The BR consists of a single bit. It can provide a serial path between the TDI input and TDO output, bypassing the BSR to reduce test access times.

# 4.2.3 BOUNDARY SCAN REGISTER (BSR)

The BSR can apply and read test patterns in parallel to or from all the digital I/O pins. The BSR is a 98 bits long shift register and is initialized and read using the instruction EXTEST or SAMPLE/PRELOAD. Each pin is related to one or more bits in the BSR. Please refer to Table-19 for details of BSR bits and their functions.

Table-19 Boundary Scan Register Description (Continued)

Bit No.	Bit Symbol	Pin Signal	Туре	Comments
				Controls pins LP[7:0].
16	PIOS	N/A	-	When '0', the pins are configured as outputs. The output values to the pins are set in POUT 7~0. When '1', the pins are high-Z. The input values to the pins are read in PIN 7~0.
17	TCLK1	TCLK1	I	
18	TDP1	TDP1	I	
19	TDN1	TDN1		
20	RCLK1	RCLK1	0	
21	RDP1	RDP1	0	
22	RDN1	RDN1	0	
23	HZEN1	N/A	-	Controls pin RDP1, RDN1 and RCLK1.  When '0', the outputs are enabled on the pins.  When '1', the pins are high-Z.
24	LOS1	LOS1	0	
25	TCLK0	TCLK0	I	
26	TDP0	TDP0	I	
27	TDN0	TDN0	I	
28	RCLK0	RCLK0	0	
29	RDP0	RDP0	0	
30	RDN0	RDN0	0	
31	HZEN0	N/A	-	Controls pin RDP0, RDN0 and RCLK0. When '0', the outputs are enabled on the pins. When '1', the pins are high-Z.
32	LOS0	LOS0	0	
33	MODE1	MODE1	I	
34	LOS3	LOS3	0	
35	RDN3	RDN3	0	
36	RDP3	RDP3	0	
37	HZEN3	N/A	-	Controls pin RDP3, RDN3 and RCLK3. When '0', the outputs are enabled on the pins. When '1', the pins are high-Z.
38	RCLK3	RCLK3	0	
39	TDN3	TDN3	I	
40	TDP3	TDP3	I	
41	TCLK3	TCLK3	I	
42	LOS2	LOS2	0	
43	RDN2	RDN2	0	
44	RDP2	RDP2	0	
45	HZEN2	N/A	-	Controls pin RDP2, RDN2 and RCLK2. When '0', the outputs are enabled on the pins. When '1', the pins are high-Z.
46	RCLK2	RCLK2	0	
47	TDN2	TDN2	I	
48	TDP2	TDP2	I	
49	TCLK2	TCLK2	I	
50	INT	ĪNT	0	
51	ACK	ACK	0	
52	SDORDYS	N/A	-	Control pin ACK. When '0', the output is enabled on pin ACK. When '1', the pin is high-Z.
53	WRB	DS	I	
54	RDB	R/W	I	
55	ALE	ALE	I	
56	CSB	CS		

**Table-19 Boundary Scan Register Description (Continued)** 

Bit No.	Bit Symbol	Pin Signal	Туре	Comments
57	MODE0	MODE0	I	
58	TCLK5	TCLK5		
59	TDP5	TDP5	[	
60	TDN5	TDN5		
61	RCLK5	RCLK5	0	
62	RDP5	RDP5	0	
63	RDN5	RDN5	0	
64	HZEN5	N/A	-	Controls pin RDP5, RDN5 and RCLK5. When '0', the outputs are enabled on the pins. When '1', the pins are high-Z.
65	LOS5	LOS5	0	
66	TCLK4	TCLK4	I	
67	TDP4	TDP4	I	
68	TDN4	TDN4	I	
69	RCLK4	RCLK4	0	
70	RDP4	RDP4	0	
71	RDN4	RDN4	0	
72	HZEN4	N/A	-	Controls pin RDP4, RDN4 and RCLK4. When '0', the outputs are enabled on the pins. When '1', the pins are high-Z.
73	LOS4	LOS4	0	
74	OE	OE		
75	CLKE	CLKE		
76	LOS7	LOS7	0	
77	RDN7	RDN7	0	
78	RDP7	RDP7	0	
79	HZEN7	N/A	-	Controls pin RDP7, RDN7 and RCLK7. When '0', the outputs are enabled on the pins. When '1', the pins are high-Z.
80	RCLK7	RCLK7	0	
81	TDN7	TDN7		
82	TDP7	TDP7		
83	TCLK7	TCLK7		
84	LOS6	LOS6	0	
85	RDN6	RDN6	0	
86	RDP6	RDP6	0	
87	HZEN6	N/A	-	Controls pin RDP6, RDN6 and RCLK6. When '0', the outputs are enabled on the pins. When '1', the pins are high-Z.
88	RCLK6	RCLK6	0	
89	TDN6	TDN6		
90	TDP6	TDP6	I	
91	TCLK6	TCLK6	I	
92	MCLK	MCLK	I	
93	MODE2	MODE2		
94	A4	A4		
95	A3	A3		
96	A2	A2		
97	A1	A1		
98	A0	A0	Ī	

## 4.3 TEST ACCESS PORT CONTROLLER

The TAP controller is a 16-state synchronous state machine. Figure-22 shows its state diagram A description of each state follows. Note that the figure contains two main branches to access either the data or instruction registers. The value shown next to each state transition in this figure states the value present at TMS at each rising edge of TCK. Refer to Table-20 for details of the state description.

### **Table-20 TAP Controller State Description**

State	Description
Test Logic Reset	In this state, the test logic is disabled. The device is set to normal operation. During initialization, the device initializes the instruction register with the IDCODE instruction.  Regardless of the original state of the controller, the controller enters the Test-Logic-Reset state when the TMS input is held high for at least 5 rising edges of TCK. The controller remains in this state while TMS is high. The device processor automatically enters this state at power-up.
Run-Test/Idle	This is a controller state between scan operations. Once in this state, the controller remains in the state as long as TMS is held low. The instruction register and all test data registers retain their previous state. When TMS is high and a rising edge is applied to TCK, the controller moves to the Select-DR state.
Select-DR-Scan	This is a temporary controller state and the instruction does not change in this state. The test data register selected by the current instruction retains its previous state. If TMS is held low and a rising edge is applied to TCK when in this state, the controller moves into the Capture-DR state and a scan sequence for the selected test data register is initiated. If TMS is held high and a rising edge applied to TCK, the controller moves to the Select-IR-Scan state.
Capture-DR	In this state, the Boundary Scan Register captures input pin data if the current instruction is EXTEST or SAMPLE/PRELOAD. The instruction does not change in this state. The other test data registers, which do not have parallel input, are not changed. When the TAP controller is in this state and a rising edge is applied to TCK, the controller enters the Exit1-DR state if TMS is high or the Shift-DR state if TMS is low.
Shift-DR	In this controller state, the test data register connected between TDI and TDO as a result of the current instruction shifts data on stage toward its serial output on each rising edge of TCK. The instruction does not change in this state. When the TAP controller is in this state and a rising edge is applied to TCK, the controller enters the Exit1-DR state if TMS is high or remains in the Shift-DR state if TMS is low.
Exit1-DR	This is a temporary state. While in this state, if TMS is held high, a rising edge applied to TCK causes the controller to enter the Update-DR state, which terminates the scanning process. If TMS is held low and a rising edge is applied to TCK, the controller enters the Pause-DR state. The test data register selected by the current instruction retains its previous value and the instruction does not change during this state.
Pause-DR	The pause state allows the test controller to temporarily halt the shifting of data through the test data register in the serial path between TDI and TDO. For example, this state could be used to allow the tester to reload its pin memory from disk during application of a long test sequence. The test data register selected by the current instruction retains its previous value and the instruction does not change during this state. The controller remains in this state as long as TMS is low. When TMS goes high and a rising edge is applied to TCK, the controller moves to the Exit2-DR state.
Exit2-DR	This is a temporary state. While in this state, if TMS is held high, a rising edge applied to TCK causes the controller to enter the Update-DR state, which terminates the scanning process. If TMS is held low and a rising edge is applied to TCK, the controller enters the Shift-DR state. The test data register selected by the current instruction retains its previous value and the instruction does not change during this state.
Update-DR	The Boundary Scan Register is provided with a latched parallel output to prevent changes while data is shifted in response to the EXTEST and SAMPLE/PRELOAD instructions. When the TAP controller is in this state and the Boundary Scan Register is selected, data is latched into the parallel output of this register from the shift-register path on the falling edge of TCK. The data held at the latched parallel output changes only in this state. All shift-register stages in the test data register selected by the current instruction retain their previous value and the instruction does not change during this state.
Select-IR-Scan	This is a temporary controller state. The test data register selected by the current instruction retains its previous state. If TMS is held low and a rising edge is applied to TCK when in this state, the controller moves into the Capture-IR state, and a scan sequence for the instruction register is initiated. If TMS is held high and a rising edge is applied to TCK, the controller moves to the Test-Logic-Reset state. The instruction does not change during this state.
Capture-IR	In this controller state, the shift register contained in the instruction register loads a fixed value of '100' on the rising edge of TCK. This supports fault-isolation of the board-level serial test data path. Data registers selected by the current instruction retain their value and the instruction does not change during this state. When the controller is in this state and a rising edge is applied to TCK, the controller enters the Exit1-IR state if TMS is held high, or the Shift-IR state if TMS is held low.
Shift-IR	In this state, the shift register contained in the instruction register is connected between TDI and TDO and shifts data one stage towards its serial output on each rising edge of TCK. The test data register selected by the current instruction retains its previous value and the instruction does not change during this state. When the controller is in this state and a rising edge is applied to TCK, the controller enters the Exit1-IR state if TMS is held high, or remains in the Shift-IR state if TMS is held low.

## **Table-20 TAP Controller State Description (Continued)**

State	Description
Exit1-IR	This is a temporary state. While in this state, if TMS is held high, a rising edge applied to TCK causes the controller to enter the Update-IR state, which terminates the scanning process. If TMS is held low and a rising edge is applied to TCK, the controller enters the Pause-IR state. The test data register selected by the current instruction retains its previous value and the instruction does not change during this state.
Pause-IR	The pause state allows the test controller to temporarily halt the shifting of data through the instruction register. The test data register selected by the current instruction retains its previous value and the instruction does not change during this state. The controller remains in this state as long as TMS is low. When TMS goes high and a rising edge is applied to TCK, the controller moves to the Exit2-IR state.
Exit2-IR	This is a temporary state. While in this state, if TMS is held high, a rising edge applied to TCK causes the controller to enter the Update-IR state, which terminates the scanning process. If TMS is held low and a rising edge is applied to TCK, the controller enters the Shift-IR state. The test data register selected by the current instruction retains its previous value and the instruction does not change during this state.
Update-IR	The instruction shifted into the instruction register is latched into the parallel output from the shift-register path on the falling edge of TCK. When the new instruction has been latched, it becomes the current instruction. The test data registers selected by the current instruction retain their previous value.

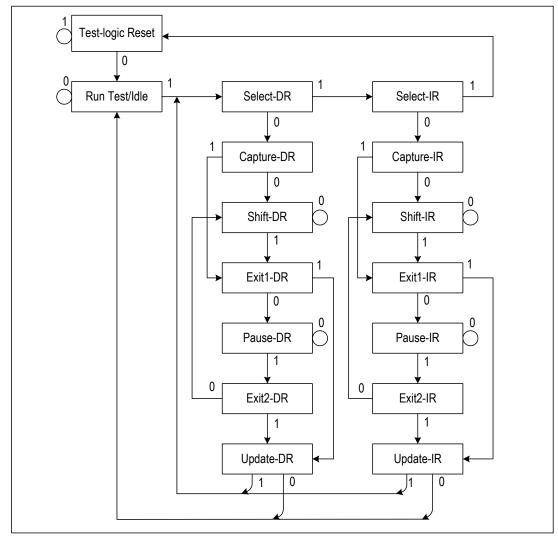


Figure-22 JTAG State Diagram

## **ABSOLUTE MAXIMUM RATING**

Symbol	Parameter	Min	Max	Unit
VDDA, VDDD	Core Power Supply	-0.5	4.0	V
VDDIO0, VDDIO1	I/O Power Supply	-0.5	4.0	V
VDDT0-7	Transmit Power Supply	-0.5	7.0	V
	Input Voltage, any digital pin	GND-0.5	5.5	V
Vin	Input Voltage <sup>(1)</sup> , RTIPn pins and RRINGn pins	GND-0.5	VDDA+ 0.5 VDDD+ 0.5	V V
	ESD Voltage, any pin <sup>(2)</sup>	2000		V
	Transient Latch-up Current, any pin		100	mA
lin	Input Current, any digital pin <sup>(3)</sup>	-10	10	mA
	DC Input Current, any analog pin <sup>(3)</sup>		±100	mA
Pd	Maximum Power Dissipation in package		1.6	W
Tc	Case Temperature		120	°C
Ts	Storage Temperature	-65	+150	°C

**CAUTION**: Exceeding these values may cause permanent damage. Functional operation under these conditions is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## RECOMMENDED OPERATING CONDITIONS

Symbol	Parameter	Min	Тур	Max	Unit
VDDA, VDDD	Core Power Supply	3.13	3.3	3.47	V
VDDIO	I/O Power Supply	3.13	3.3	3.47	V
VDDT	Transmitter Supply				
	3.3 V	3.13	3.3	3.47	V
	5 V	4.75	5.0	5.25	V
T <sub>A</sub>	Ambient Operating Temperature	-40	25	85	°C
R <sub>L</sub>	Output load at TTIPn pins and TRINGn pins	25			W
I <sub>VDD</sub>	Average Core Power Supply Current <sup>(1)</sup>		40	60	mA
I <sub>VDDIO</sub>	I/O Power Supply Current <sup>(2)</sup>		15	25	mA
I <sub>VDDT</sub>	Average transmitter power supply current, E1 mode <sup>(1), (3)</sup>				
	75 $\Omega$ 50% ones density data:			125	mA
	100% ones density data:			220	mA
	120 $\Omega$ 50% ones density data:			100	mA
1.00	100% ones density data:			200	mA

<sup>1</sup> Maximum power and current consumption over the full operating temperature and power supply voltage range. Includes all channels.

<sup>1.</sup> Referenced to ground

<sup>&</sup>lt;sup>2.</sup> Human body model

<sup>3.</sup> Constant input current

<sup>&</sup>lt;sup>2.</sup> Digital output is driving 50 pF load, digital input is within 10% of the supply rails.

<sup>&</sup>lt;sup>3.</sup> Power consumption includes power absorbed by line load and external transmitter components.

## **POWER CONSUMPTION**

Symbol	Parameter	Min	Тур	Max <sup>(1)(2)</sup>	Unit
	E1, 3.3 V, 75 Ω Load				
	50% ones density data:	-	612	-	mW
	100% ones density data:	-	1050	1125	mW
	E1, 3.3 V, 120 Ω Load				
	50% ones density data:	-	526	-	mW
	100% ones density data:	-	880	940	mW
	E1, 5.0 V, 75 Ω Load				
	50% ones density data:	-	835	-	mW
	100% ones density data:	-	1510	1610	mW
	E1, 5.0 V, 120 Ω Load				
	50% ones density data:	-	710	-	mW
	100% ones density data:	-	1240	1330	mW

<sup>1</sup> Maximum power and current consumption over the full operating temperature and power supply voltage range. Includes all channels.

## **DC CHARACTERISTICS**

Symbol	Parameter	Min	Тур	Max	Unit
V <sub>IL</sub>	Input Low Level Voltage				
	MODE2, JAS and LPn pins			$\frac{1}{3}$ VDDIO-0.2	V
	All other digital inputs pins			0.8	V
V <sub>IM</sub>	Input Mid Level Voltage				
	MODE2, JAS and LPn pins	$\frac{1}{3}$ VDDIO+0.2	$\frac{1}{2}$ VDDIO	$\frac{2}{3}$ VDDIO-0.2	V
V <sub>IH</sub>	Input High Voltage				
	MODE2, JAS and LPn pins	$\frac{2}{3}$ VDDIO+ 0.2			V
	All other digital inputs pins	2.0			V
$V_{OL}$	Output Low level Voltage <sup>(1)</sup> (lout = 1.6 mA)			0.4	V
V <sub>OH</sub>	Output High level Voltage <sup>(1)</sup> (lout = 400 μA)	2.4		VDDIO	V
$V_{MA}$	Analog Input Quiescent Voltage (RTIPn/RRINGn pin while floating)	1.33	1.4	1.47	V
I <sub>H</sub>	Input High Level Current (MODE2, JAS and LPn pin)			50	μΑ
ΙL	Input Low Level Current (MODE2, JAS and LPn pin)			50	μΑ
l <sub>l</sub>	Input Leakage Current				
	TMS, TDI and TRST pins			50	μA
	All other digital input pins	-10		10	μΑ
$I_{ZL}$	High-Z Leakage Current	-10		10	μΑ
Z <sub>OH</sub>	Output High-Z on TTIPn pins and TRINGn pins	150	<u> </u>		kΩ

<sup>1.</sup> Output drivers will output CMOS logic levels into CMOS loads.

<sup>&</sup>lt;sup>2.</sup> Power consumption includes power absorbed by line load and external transmitter components.

# TRANSMITTER CHARACTERISTICS

Symbol		Parameter		Тур	Max	Unit
V <sub>o-p</sub>	Output Pulse Amplitudes <sup>(1)</sup> 75 $\Omega$ load 120 $\Omega$ load		2.14 2.7	2.37 3.0	2.6 3.3	V V
V <sub>o-s</sub>	Zero (space) Level 75 $\Omega$ load 120 $\Omega$ load		-0.237 -0.3		0.237 0.3	V
	Transmit Amplitude Variation with supply	1	-1		+1	%
	Difference between pulse sequences for	17 consecutive pulses			200	mV
T <sub>PW</sub>	Output Pulse Width at 50% of nominal a	mplitude	232	244	256	ns
	Ratio of the amplitudes of Positive and N	egative Pulses at the center of the pulse interval	0.95		1.05	
RTX	Transmit Return Loss <sup>(2)</sup>		•			
	75 Ω	51 kHz – 102 kHz 102 kHz – 2.048 MHz 2.048 MHz – 3.072 MHz	15 15 15			dB dB dB
	120 Ω	51 kHz – 102 kHz 102 kHz – 2.048 MHz 2.048 MHz – 3.072 MHz	15 15 15			dB dB dB
JTX <sub>P-P</sub>	Intrinsic Transmit Jitter (TCLK is jitter fre	e, JA enabled)				
	20 Hz – 100 kHz			0.050		U.I.
Td	Transmit Path Delay (JA is disabled)					
	Single Rail Dual Rail			8 3		U.I. U.I.
I <sub>SC</sub>	Line Short Circuit Current <sup>(3)</sup>			180		mAp

<sup>1.</sup> Measured at the line output ports

<sup>&</sup>lt;sup>2.</sup> Test at IDT82V2058 evaluation board

<sup>3.</sup> Measured on device, between TTIPn and TRINGn

## **RECEIVER CHARACTERISTICS**

Symbol	Parameter	Min	Тур	Max	Unit
ATT	Permissible Cable Attenuation (@1024 kHz)			15	dB
IA	Input Amplitude	0.1		0.9	Vp
SIR	Signal to Interference Ratio Margin <sup>(1)</sup>	-15			dB
SRE	Data Decision Threshold (refer to peak input voltage)		50		%
	Data Slicer Threshold		150		mV
	Analog Loss Of Signal <sup>(2)</sup> Declare/Clear:	120/150	200/250	280/350	mVp
	Allowable consecutive zeros before LOS G.775: ETSI 300 233:		32 2048		
	LOS Reset Clock Recovery Mode	12.5			% ones
JRX <sub>p-p</sub>	Peak to Peak Intrinsic Receive Jitter (JA disabled)			0.0625	U.I.
JTRX	Jitter Tolerance				
	1 Hz – 20 Hz	18.0			U.I.
	20 Hz – 2.4 kHz	1.5			U.I.
	18 kHz – 100 kHz	0.2			U.I.
ZDM	Receiver Differential Input Impedance		120		kΩ
ZCM	Receiver Common Mode Input Impedance to GND	10			kΩ
RRX	Receive Return Loss 51 kHz – 102 kHz 102 kHz – 2.048 MHz 2.048 MHz – 3.072 MHz	20 20 20			dB dB dB
	Receive Path Delay  Dual rail  Single rail		3 8		U.I. U.I.

<sup>1.</sup> Per G.703, O.151 @ 6 dB cable attenuation

## **JITTER ATTENUATOR CHARACTERISTICS**

Symbol	Parameter	Min	Тур	Max	Unit
f <sub>-3dB</sub>	Jitter Transfer Function Corner Frequency (–3 dB)				
	Host mode: 32/64 bit FIFO  JABW = 0:  JABW = 1:  Hardware mode		1.7 6.6 1.7		Hz Hz Hz
	Jitter Attenuator <sup>(1)</sup>				
	@ 3 Hz @ 40 Hz @ 400 Hz @ 100 kHz	-0.5 -0.5 +19.5 +19.5			dB dB dB dB
td	Jitter Attenuator Latency Delay 32 bit FIFO: 64 bit FIFO:		16 32		U.I. U.I.
	Input Jitter Tolerance before FIFO Overflow Or Underflow 32 bit FIFO: 64 bit FIFO:		28 56		U.I. U.I.
	Output Jitter in Remote Loopback <sup>(2)</sup>			0.11	U.I.

<sup>&</sup>lt;sup>1.</sup> Per G.736, see Figure-38 on page 52.

<sup>&</sup>lt;sup>2.</sup> Measured on device, between RTIP and RRING, all ones signal.

<sup>&</sup>lt;sup>2.</sup> Per ETSI CTR12/13 output jitter.

## TRANSCEIVER TIMING CHARACTERISTICS

Symbol	Parameter	Min	Тур	Max	Unit
	MCLK Frequency		2.048		MHz
	MCLK Tolerance	-100		100	ppm
	MCLK Duty Cycle	40		60	%
Transmit path	•	•	1	1	
	TCLK Frequency		2.048		MHz
	TCLK Tolerance	-50		+50	ppm
	TCLK Duty Cycle	10		90	%
t1	Transmit Data Setup Time	40			ns
t2	Transmit Data Hold Time	40			ns
	Delay Time of OE Low to Driver High-Z			1	μs
	Delay Time of TCLK Low to Driver High-Z	40	44	48	μs
Receive path		•	1	1	
	Clock Recovery Capture Range <sup>(1)</sup>		± 80		ppm
	RCLK Duty Cycle <sup>(2)</sup>	40	50	60	%
t4	RCLK Pulse Width <sup>(2)</sup>	457	488	519	ns
t5	RCLK Pulse Width Low Time	203	244	285	ns
t6	RCLK Pulse Width High Time	203	244	285	ns
	Rise/Fall Time <sup>(3)</sup>	5		30	ns
t7	Receive Data Setup Time	200	244		ns
t8	Receive Data Hold Time	200	244		ns
t9	RDPn/RDNn Pulse Width (MCLK = High) <sup>(4)</sup>	200	244		ns

<sup>1.</sup> Relative to nominal frequency, MCLK =  $\pm$  100 ppm

<sup>&</sup>lt;sup>2.</sup> RCLK duty cycle widths will vary depending on extent of received pulse jitter displacement. Maximum and minimum RCLK duty cycles are for worst case jitter conditions (0.2 UI displacement for E1 per ITU G.823).

<sup>&</sup>lt;sup>3.</sup> For all digital outputs. C load = 15 pF

<sup>&</sup>lt;sup>4.</sup> Clock recovery is disabled in this mode.

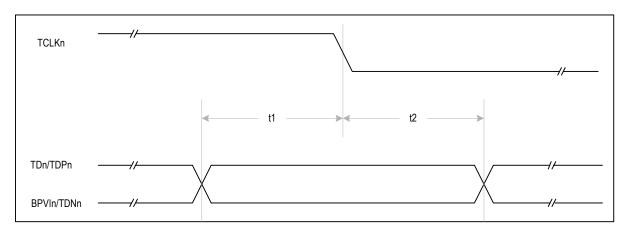


Figure-23 Transmit System Interface Timing

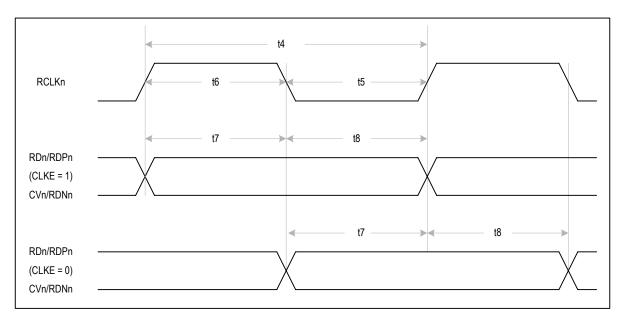


Figure-24 Receive System Interface Timing

# **JTAG TIMING CHARACTERISTICS**

Symbol	Parameter	Min	Тур	Max	Unit	Comments
t1	TCK Period	200			ns	
t2	TMS to TCK setup Time TDI to TCK Setup Time	50			ns	
t3	TCK to TMS Hold Time TCK to TDI Hold Time	50			ns	
t4	TCK to TDO Delay Time			100	ns	

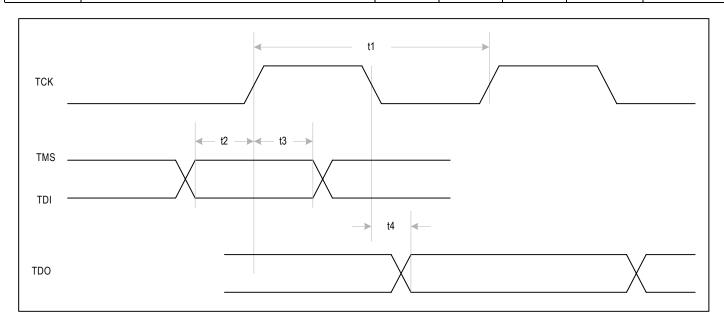


Figure-25 JTAG Interface Timing

# PARALLEL HOST INTERFACE TIMING CHARACTERISTICS

#### INTEL MODE READ TIMING CHARACTERISTICS

Symbol	Parameter	Min	Тур	Max	Unit	Comments
t1	Active RD Pulse Width	90			ns	(1)
t2	Active CS to Active RD Setup Time	0			ns	
t3	Inactive RD to Inactive CS Hold Time	0			ns	
t4	Valid Address to Inactive ALE Setup Time (in Multiplexed Mode)	5			ns	
t5	Invalid RD to Address Hold Time (in Non-Multiplexed Mode)	0			ns	
t6	Active RD to Data Output Enable Time	7.5		15	ns	
t7	Inactive RD to Data High-Z Delay Time	7.5		15	ns	
t8	Active CS to RDY delay time	6		12	ns	
t9	Inactive CS to RDY High-Z Delay Time	6		12	ns	
t10	Inactive RD to Inactive INT Delay Time			20	ns	
t11	Address Latch Enable Pulse Width (in Multiplexed Mode)	10			ns	
t12	Address Latch Enable to RD Setup Time (in Multiplexed Mode)	0			ns	
t13	Address Setup time to Valid Data Time (in Non-Multiplexed Mode)	18		32	ns	
t14	Inactive RD to Active RDY Delay Time	10		15	ns	
t15	Active RD to Active RDY Delay Time	30		85	ns	
t16	Inactive ALE to Address Hold Time (in Multiplexed Mode)	5			ns	

<sup>&</sup>lt;sup>1</sup> The t1 is determined by the start time of the valid data when the RDY signal is not used.

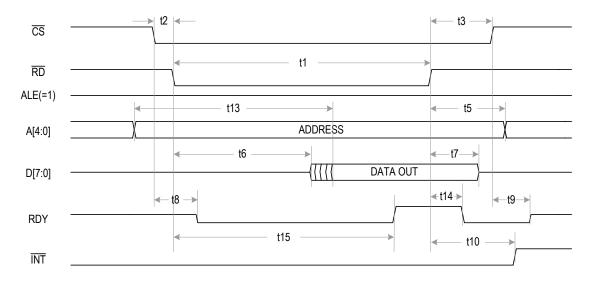


Figure-26 Non-Multiplexed Intel Mode Read Timing

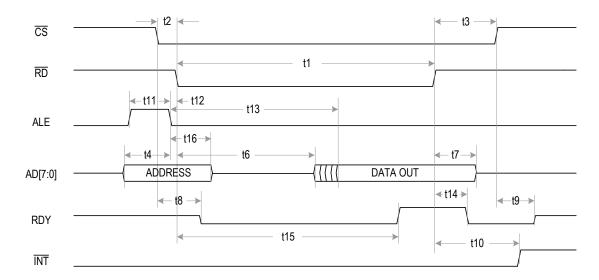


Figure-27 Multiplexed Intel Mode Read Timing

#### **INTEL MODE WRITE TIMING CHARACTERISTICS**

Symbol	Parameter	Min	Тур	Max	Unit	Comments
t1	Active WR Pulse Width	90			ns	(1)
t2	Active CS to Active WR Setup Time	0			ns	
t3	Inactive WR to Inactive CS Hold Time	0			ns	
t4	Valid Address to Latch Enable Setup Time (in Multiplexed Mode)	5			ns	
t5	Invalid WR to Address Hold Time (in Non-Multiplexed Mode)	2			ns	
t6	Valid Data to Inactive WR Setup Time	5			ns	
t7	Inactive WR to Data Hold Time	10			ns	
t8	Active CS to Inactive RDY Delay Time	6		12	ns	
t9	Active WR to Active RDY Delay Time	30		85	ns	
t10	Inactive WR to Inactive RDY Delay Time	10		15	ns	
t11	Invalid CS to RDY High-Z Delay Time	6		12	ns	
t12	Address Latch Enable Pulse Width (in Multiplexed Mode)	10			ns	
t13	Inactive ALE to WR Setup Time (in Multiplexed Mode)	0			ns	
t14	Inactive ALE to Address hold time (in Multiplexed Mode)	5			ns	
t15	Address setup time to Inactive WR time (in Non-Multiplexed Mode)	5			ns	_

<sup>1.</sup> The t1 can be 15 ns when RDY signal is not used.

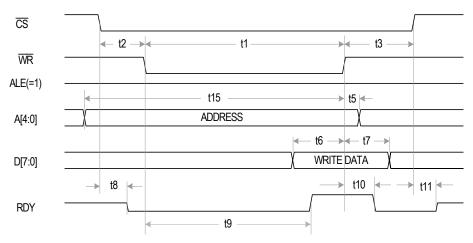


Figure-28 Non-Multiplexed Intel Mode Write Timing

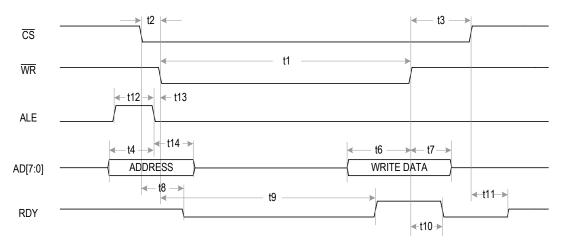


Figure-29 Multiplexed Intel Mode Write Timing

#### **MOTOROLA MODE READ TIMING CHARACTERISTICS**

Symbol	Parameter	Min	Тур	Max	Unit	Comments
t1	Active DS Pulse Width	90			ns	(1)
t2	Active CS to Active DS Setup Time	0			ns	
t3	Inactive DS to Inactive CS Hold Time	0			ns	
t4	Valid R/W to Active DS Setup Time	0			ns	
t5	Inactive $\overline{\text{DS}}$ to R/W Hold Time	0.5			ns	
t6	Valid Address to Active DS Setup Time (in Non-Multiplexed Mode)	5			ns	
t7	Active DS to Address Hold Time (in Non-Multiplexed Mode)	10			ns	
t8	Active DS to Data Valid Delay Time (in Non-Multiplexed Mode)	20		35	ns	
t9	Active DS to Data Output Enable Time	7.5		15	ns	
t10	Inactive $\overline{\text{DS}}$ to Data High-Z Delay Time	7.5		15	ns	
t11	Active DS to Active ACK Delay Time	30		85	ns	
t12	Inactive DS to Inactive ACK Delay Time	10		15	ns	
t13	Inactive $\overline{\text{DS}}$ to Invalid $\overline{\text{INT}}$ Delay Time			20	ns	
t14	Active AS to Active DS Setup Time (in Multiplexed Mode)	5			ns	

<sup>&</sup>lt;sup>1</sup> The t1 is determined by the start time of the valid data when the ACK signal is not used.

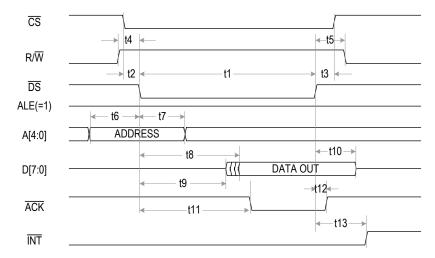


Figure-30 Non-Multiplexed Motorola Mode Read Timing

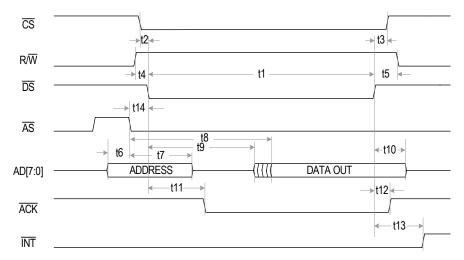


Figure-31 Multiplexed Motorola Mode Read Timing

#### **MOTOROLA MODE WRITE TIMING CHARACTERISTICS**

Symbol	Parameter	Min	Тур	Max	Unit	Comments
t1	Active DS Pulse Width	90			ns	(1)
t2	Active CS to Active DS Setup Time	0			ns	
t3	Inactive $\overline{\text{DS}}$ to Inactive $\overline{\text{CS}}$ Hold Time	0			ns	
t4	Valid R/W to Active DS Setup Time	10			ns	
t5	Inactive $\overline{\rm DS}$ to R/ $\overline{\rm W}$ Hold Time	0			ns	
t6	Valid Address to Active DS Setup Time (in Non-Multiplexed Mode)	10			ns	
t7	Valid DS to Address Hold Time (in Non-Multiplexed Mode)	10			ns	
t8	Valid Data to Inactive DS Setup Time	5			ns	
t9	Inactive DS to Data Hold Time	10			ns	
t10	Active DS to Active ACK Delay Time	30		85	ns	
t11	Inactive DS to Inactive ACK Delay Time	10		15	ns	
t12	Active AS to Active DS (in Multiplexed Mode)	0			ns	
t13	Inactive DS to Inactive AS Hold Time ( in Multiplexed Mode)	15			ns	

<sup>1.</sup> The t1 can be 15ns when the ACK signal is not used.

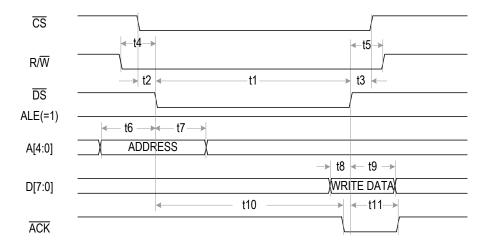


Figure-32 Non-Multiplexed Motorola Mode Write Timing

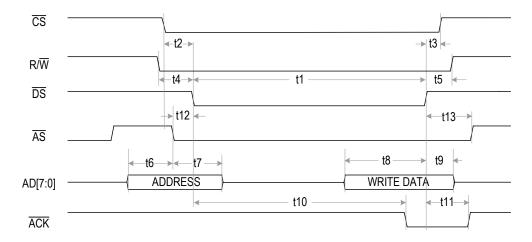


Figure-33 Multiplexed Motorola Mode Writing Timing

#### SERIAL HOST INTERFACE TIMING CHARACTERISTICS

Symbol	Parameter	Min	Тур	Max	Unit	Comments
t1	SCLK High Time	25			ns	
t2	SCLK Low Time	25			ns	
t3	Active CS to SCLK Setup Time	10			ns	
t4	Last SCLK Hold Time to Inactive CS Time	50			ns	
t5	CS Idle Time	50			ns	
t6	SDI to SCLK Setup Time	5			ns	
t7	SCLK to SDI Hold Time	5			ns	
t8	Rise/Fall Time (any pin)			100	ns	
t9	SCLK Rise and Fall Time			50	ns	
t10	SCLK to SDO Valid Delay Time		25	35	ns	Load = 50 pF
t11	SCLK Falling Edge to SDO High-Z Hold Time (CLKE = 0) or $\overline{\text{CS}}$ Rising Edge to SDO High-Z Hold Time (CLKE = 1)		100		ns	

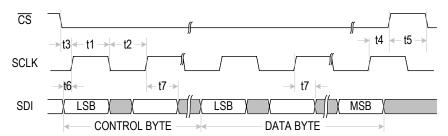


Figure-34 Serial Interface Write Timing

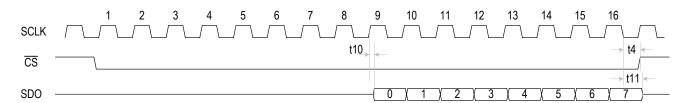


Figure-35 Serial Interface Read Timing with CLKE = 0

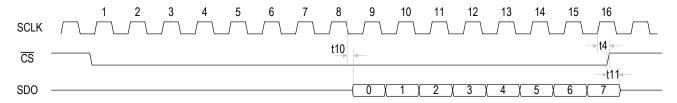
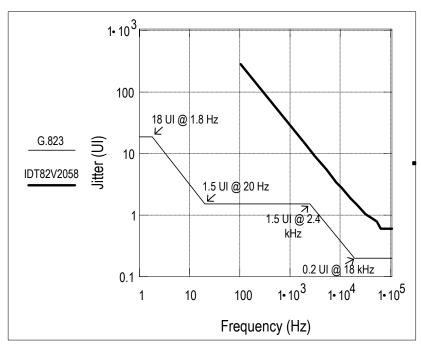


Figure-36 Serial Interface Read Timing with CLKE = 1

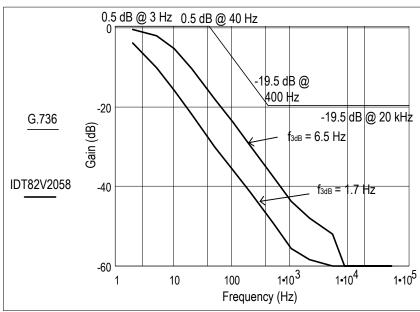
## **JITTER TOLERANCE PERFORMANCE**



Test condition: PRBS 2^15-1; Line code rule HDB3 is used.

Figure-37 Jitter Tolerance Performance

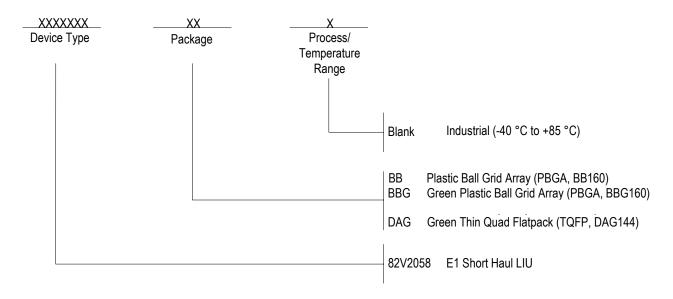
# **JITTER TRANSFER PERFORMANCE**



Test condition: PRBS 2^15-1; Line code rule HDB3 is used.

Figure-38 Jitter Transfer Performance

### **ORDERING INFORMATION**



## **DATASHEET DOCUMENT HISTORY**

11/04/2001	pgs. 2, 3, 10, 17
11/20/2001	pgs. 5, 6, 11, 13, 16, 17, 24, 26, 31, 38, 39, 40, 50
11/28/2001	pgs. 5, 24, 26, 31
11/29/2001	pgs. 5
12/05/2001	pgs. 9
01/24/2002	pgs. 2, 3, 9, 14, 39, 40
02/21/2002	pgs. 14, 16, 41
03/25/2002	pgs. 1, 2, 52
04/17/2002	pgs. 17
05/07/2002	pgs. 14, 44, 45, 48
01/15/2003	pgs. 1, 52
04/12/2005	pgs. 1, 5 to 8, 10, 11, 14, 15, 18, 19, 29, 30, 40 to 43, 47 to 53
09/14/2009	pg. 40
01/21/2010	pg. 8
10/10/2014	pg. removed DA package information



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